

Distortion-transmission trade-off in real-time transmission of Markov sources

Jhelum Chakravorty and Aditya Mahajan

Abstract

The problem of optimal real-time transmission of a Markov source under constraints on the expected number of transmissions is considered, both for the discounted and long term average cases. This setup is motivated by applications where transmission is sporadic and the cost of switching on the radio and transmitting is significantly more important than the size of the transmitted data packet. For this model, we characterize the distortion-transmission function, i.e., the minimum expected distortion that can be achieved when the expected number of transmissions is less than or equal to a particular value. In particular, we show that the distortion-transmission function is a piecewise linear, convex, and decreasing function. We also give an explicit characterization of each vertex of the piecewise linear function.

To prove the results, the optimization problem is cast as a decentralized constrained stochastic control problem. We first consider the Lagrange relaxation of the constrained problem and identify the structure of optimal transmission and estimation strategies. In particular, we show that the optimal transmission is of a threshold type. Using these structural results, we obtain dynamic programs for the Lagrange relaxations. We identify the performance of an arbitrary threshold-type transmission strategy and use the idea of calibration from multi-armed bandits to determine the optimal transmission strategy for the Lagrange relaxation. Finally, we show that the optimal strategy for the constrained setup is a randomized strategy that randomizes between two deterministic strategies that differ only at one state. By evaluating the performance of these strategies, we determine the shape of the distortion-transmission function. These results are illustrated using an example of transmitting a birth-death Markov source.

Index Terms

Real-time communication, remote estimation, team-theory, constrained Markov decision processes.

I. INTRODUCTION

A. Motivation and literature overview

In many applications such as networked control systems, sensor and surveillance networks, and transportation networks, etc., data must be transmitted sequentially from one node to another under a strict delay deadline. In many of such *real-time* communication systems, the transmitter is a battery powered device that transmits over a wireless packet-switched network; the cost of switching on the radio and transmitting a packet is significantly more important than the size of the data packet. Therefore, the transmitter does not transmit all the time; but when it does transmit, the transmitted packet is as big as needed to communicate the current source realization. In this paper, we characterize a fundamental trade-off between the real-time (i.e. zero-delay) distortion and the average number of transmissions in such systems.

In particular, we consider a transmitter that observes a first-order Markov source. At each time instant, based on the current source symbol and the history of its past decisions, the transmitter determines whether or not to transmit the current source symbol. If the transmitter does not transmit, the receiver must estimate the source symbol using the previously transmitted values. A per-step distortion function measures the fidelity of estimation. We are interested in characterizing the optimal transmission and estimation strategies that minimize the expected distortion over an infinite horizon under a constraint on the expected number of transmissions.

The communication system described above is similar to the classical information theory setup. In particular, it may be viewed as minimizing the average distortion while transmitting over a channel under an average-power constraint. However, unlike the classical information theory setup, the source reconstruction must be done in real-time (i.e. with zero delay). Due to this real-time constraint on source reconstruction, traditional information theoretic approach does not apply.

Two approaches have been used in the literature to investigate real-time or zero-delay communication. The first approach considers coding of individual sequences [1]–[4]; the second approach considers coding of Markov sources [5]–[10]. The model presented above fits with the latter approach. In particular, it may be viewed as real-time transmission over a noiseless channel with input cost. In most of the results in the literature on real-time coding of Markov sources, the focus has been on identifying sufficient statistics (or information states) at the transmitter and the receiver; for some of the models, a dynamic programming decomposition has also been derived. However, very little is known about the solution of these dynamic programs.

This paper was presented in part in the Proceedings of the 52nd Annual Allerton Conference on Communication, Control, and Computing, 2014 and the Proceedings of the 53rd Conference on Decision and Control, 2014.

The authors are with the Department of Electrical and Computer Engineering, McGill University, QC, Canada. Email: jhelum.chakravorty@mail.mcgill.ca, aditya.mahajan@mcgill.ca.

This work was supported in part by Fonds de recherche du Québec – Nature et technologies (FRQNT) Team Grant PR-173396.

The communication system described above is much simpler than the general real-time communication setup due to the following feature: whenever the transmitter transmits, it sends the current realization of the source to the receiver. These transmitted events *reset* the system. In addition, we impose certain symmetry assumptions on the model, which ensure that there is a single reset state. We exploit these special features to identify an analytic solution to the dynamic program corresponding to the above communication system. In particular, we show that threshold-based strategies are optimal at the transmitter; the optimal transmission strategy randomizes between two threshold-based strategies; the randomization takes place only at one state.

Several variations of the communication system described above have been considered in the literature. The most closely related models are [11]–[15] which are summarized below. Other related work includes censoring sensors [16], [17] (where a sensor takes a measurement and decides whether to transmit it or not; in the context of sequential hypothesis testing), estimation with measurement cost [18]–[20] (where the receiver decides when the sensor should transmit), sensor sleep scheduling [21]–[24] (where the sensor is allowed to sleep for a pre-specified amount of time); and event-based communication [25]–[27] (where the sensor transmits when a certain event takes place). We contrast our model with [11]–[15] below.

In [11], the authors considered a remote estimation problem where the sensor could communicate a finite number of times. They assumed that the sensor used a threshold strategy to decide when to communicate and determined the optimal estimation strategy and the value of the thresholds. In [12], the authors considered remote estimation of a Gauss-Markov process. They assumed a particular form of the estimator and showed that the estimation error is a sufficient statistic for the sensor.

In [13], the authors considered remote estimation of a scalar Gauss-Markov process but did not impose any assumption on the communication or estimation strategy. They used ideas from majorization theory to show that the optimal estimation strategy is Kalman-like and the optimal transmission strategy is threshold based. The results of [13] were generalized to other setups in [14] and [15]. In [14], the authors considered remote estimation of countable state Markov processes where the sensor harvests energy to communicate. Similar to the approach taken in [13], the authors used majorization theory to show that if the Markov process is driven by symmetric and unimodal noise process then the structural results of [13] continue to hold. In [15], the authors considered remote estimation of a scalar first-order autoregressive source. They used a person-by-person optimization approach to identify an iterative algorithm to compute the optimal transmission and estimation strategy. They showed that if the autoregressive process is driven by a symmetric unimodal noise process, then the iterative algorithm has a unique fixed point and the structural results of [13] continue to hold.

In all these papers [13]–[15], a dynamic program to compute the optimal thresholds was also identified.

B. Contributions

We investigate the optimal real-time transmission of a Markov source under constraints on the expected number of transmissions. Under certain symmetry assumptions on the source and the distortion function, we characterize the distortion-transmission function that describes the optimal trade-off between the expected distortion and the expected number of transmissions. In particular, we show that the distortion-transmission function is piecewise linear, convex, and decreasing.

In addition, we identify transmission and estimation strategies that achieve the minimum real-time distortion for a particular value of the expected number of transmissions. The optimal estimation strategy is deterministic, while the optimal transmission strategy possibly randomizes between two deterministic strategies that differ at only one state.

C. Notation

Throughout this paper, we use the following notation. \mathbb{Z} , $\mathbb{Z}_{\geq 0}$ and $\mathbb{Z}_{>0}$ denote the set of integers, the set of non-negative integers and the set of strictly positive integers respectively. Similarly, \mathbb{R} , $\mathbb{R}_{\geq 0}$ and $\mathbb{R}_{>0}$ denote the set of reals, the set of non-negative reals and the set of strictly positive reals respectively. Upper-case letters (e.g., X , Y) denote random variables; corresponding lower-case letters (e.g., x , y) denote their realizations. $X_{1:t}$ is a short hand notation for the vector (X_1, \dots, X_t) . Given a matrix A , A_{ij} denotes its (i, j) -th element, A_i denotes its i -th row, A^\top denotes its transpose. We index the matrices by sets of the form $\{-k, \dots, k\}$; so the indices take both positive and negative values. I_k denotes the identity matrix of dimension $k \times k$, $k \in \mathbb{Z}_{>0}$. $\mathbf{1}_k$ denotes $k \times 1$ vector of ones. $\langle v, w \rangle$ denotes the inner product between vectors v and w , $\mathbb{P}(\cdot)$ denotes the probability of an event, $\mathbb{E}[\cdot]$ denotes the expectation of a random variable, and $\mathbb{1}\{\cdot\}$ denotes the indicator function of a statement. We follow the convention of calling a sequence $\{a_k\}_{k=0}^\infty$ increasing when $a_1 \leq a_2 \leq \dots$. If all the inequalities are strict, then we call the sequence strictly increasing.

II. PROBLEM FORMULATION

A. The communication system

In this paper, we investigate the following communication setup. A transmitter causally observes a first-order Markov source $\{X_t\}_{t=0}^\infty$, where $X_t \in \mathbb{Z}$ and the initial state $X_0 = 0$. At each time, it may choose whether or not to transmit the current source observation. This decision is denoted by $U_t \in \{0, 1\}$, where $U_t = 0$ denotes no transmission and $U_t = 1$ denotes transmission. The decision to transmit is made using a *transmission strategy* $f = \{f_t\}_{t=0}^\infty$, where

$$U_t = f_t(X_{0:t}, U_{0:t-1}). \quad (1)$$

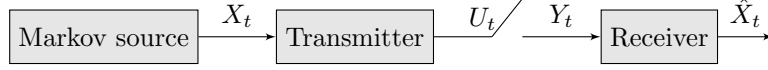


Fig. 1: A block diagram depicting the communication system considered in this paper.

We use the short-hand notation $X_{0:t}$ to denote the sequence (X_0, \dots, X_t) . Similar interpretations hold for $U_{0:t-1}$.

The transmitted symbol, which is denoted by Y_t , is given by

$$Y_t = \begin{cases} X_t, & \text{if } U_t = 1; \\ \mathfrak{E}, & \text{if } U_t = 0, \end{cases}$$

where $Y_t = \mathfrak{E}$ denotes no transmission.

The receiver causally observes $\{Y_t\}_{t=0}^\infty$ and generates a source reconstruction $\{\hat{X}_t\}_{t=0}^\infty$ (where $\hat{X}_t \in \mathbb{Z}$) in real-time using an *estimation strategy* $g = \{g_t\}_{t=0}^\infty$, i.e.,

$$\hat{X}_t = g_t(Y_{0:t}). \quad (2)$$

The fidelity of the reconstruction is measured by a per-step distortion $d(X_t - \hat{X}_t)$, where $d: \mathbb{Z} \rightarrow \mathbb{R}_{\geq 0}$.

Fig. 1 shows a communication system as described above. We impose the following assumptions on the model.

(A1) The transition matrix P of the Markov source is a Toeplitz matrix with decaying off-diagonal terms, i.e., $P_{ij} = p_{|i-j|}$, where $\{p_n\}_{n=0}^\infty$ is a decreasing non-negative sequence and $p_1 > 0$.

(A2a) The distortion function is even and increasing on $\mathbb{Z}_{\geq 0}$, i.e., for all $e \in \mathbb{Z}_{\geq 0}$

$$d(e) = d(-e) \quad \text{and} \quad d(e) \leq d(e+1).$$

(A2b) $d(0) = 0$ and $d(e) \neq 0, \forall e \neq 0$.

An example of a source and a distortion function that satisfy the above assumptions is the following:

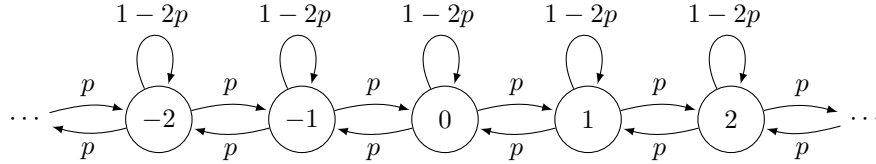


Fig. 2: A birth-death Markov chain

Example 1 Consider an aperiodic, symmetric, birth-death Markov chain defined over \mathbb{Z} as shown in Fig. 2. The transition probability matrix is given by

$$P_{ij} = \begin{cases} p, & \text{if } |i-j| = 1; \\ 1-2p, & \text{if } i=j; \\ 0, & \text{otherwise,} \end{cases}$$

where we assume that $p \in (0, \frac{1}{2})$. Let the distortion function be $d(e) = |e|$. P satisfies (A1) and $d(e)$ satisfies (A2). \square

B. The optimization problems

The objective is to choose the transmission and estimation strategies (called the *communication strategy* in short) to minimize the expected distortion under a constraint on the expected number of transmissions. We investigate two variations of this objective: the discounted setup and the long-term average setup.

1) *The discounted setup*: Given a communication strategy (f, g) and a discount factor $\beta \in (0, 1)$, let

$$D_\beta(f, g) := (1 - \beta) \mathbb{E}^{(f, g)} \left[\sum_{t=0}^{\infty} \beta^t d(X_t - \hat{X}_t) \mid X_0 = 0 \right]$$

denote the expected discounted distortion and

$$N_\beta(f, g) := (1 - \beta) \mathbb{E}^{(f, g)} \left[\sum_{t=0}^{\infty} \beta^t U_t \mid X_0 = 0 \right]$$

denote the expected discounted number of transmissions.

We are interested in the following constrained discounted cost problem: Given $\alpha \in (0, 1)$, find a strategy (f^*, g^*) such that

$$D_\beta^*(\alpha) := D_\beta(f^*, g^*) := \inf_{(f, g): N_\beta(f, g) \leq \alpha} D_\beta(f, g) \quad (\text{DIS})$$

where the infimum is taken over all history-dependent communication strategies of the form (1) and (2).

2) *The long-term average setup*: The long-term average setup is similar. Given a communication strategy (f, g) , let

$$D_1(f, g) := \limsup_{T \rightarrow \infty} \frac{1}{T} \mathbb{E}^{(f, g)} \left[\sum_{t=0}^{T-1} d(X_t - \hat{X}_t) \mid X_0 = 0 \right]$$

denote the expected long-term average distortion and

$$N_1(f, g) := \limsup_{T \rightarrow \infty} \frac{1}{T} \mathbb{E}^{(f, g)} \left[\sum_{t=0}^{T-1} U_t \mid X_0 = 0 \right]$$

denote the expected long-term average number of transmissions.

We are interested in the following constrained long-term average cost problem: Given $\alpha \in (0, 1)$, find a strategy (f^*, g^*) such that

$$D_1^*(\alpha) := D_1(f^*, g^*) := \inf_{(f, g): N_1(f, g) \leq \alpha} D_1(f, g) \quad (\text{AVG})$$

where the infimum is taken over all history-dependent communication strategies of the form (1) and (2).

C. The main result

Although the solution approach and proof techniques for Problem (DIS) and (AVG) are different, for notational convenience, we use the unified notation D_β and N_β for $\beta \in (0, 1]$ to refer to both of them.

The function $D_\beta^*(\alpha)$, $\beta \in (0, 1]$ represents the minimum expected distortion that can be achieved when the expected number of transmissions are less than or equal to α . It is analogous to the distortion-rate function in classical Information Theory; for that reason, we call it the *distortion-transmission function*.

In general, $D_\beta^*(\alpha)$ is convex and decreasing in α . This is for the following reasons. $D_\beta^*(\alpha)$ is the solution to a constrained optimization problem and the constraint set $\{(f, g) : N_\beta(f, g) \leq \alpha\}$ increases with α . Hence, $D_\beta^*(\alpha)$ decreases with α . To see that $D_\beta^*(\alpha)$ is convex in α , consider $\alpha_1 < \alpha < \alpha_2$ and suppose (f_1, g_1) and (f_2, g_2) are optimal policies for α_1 and α_2 respectively. Let $\theta = (\alpha - \alpha_1)/(\alpha_2 - \alpha_1)$ and (f, g) be a mixed strategy that picks (f_1, g_1) with probability θ and (f_2, g_2) with probability $(1 - \theta)$ (Note that the randomization is done only at the start of communication). Then $N_\beta(f, g) = \alpha$. Hence $D_\beta^*(\alpha) \leq D_\beta(f, g) = \theta D_\beta(f_1, g_1) + (1 - \theta) D_\beta(f_2, g_2)$. Hence $D_\beta^*(\alpha)$ is convex. In addition, it can be shown that $\lim_{\alpha \rightarrow 0} D_\beta^*(\alpha) = \infty$ ¹ and $\lim_{\alpha \rightarrow 1} D_\beta^*(\alpha) = 0$.

In this paper, we characterize the shape of $D_\beta^*(\alpha)$ for a class of Markov sources and distortion functions (those that satisfy (A1) and (A2)). In particular, we show that $D_\beta^*(\alpha)$ is piecewise linear (in addition to being convex and decreasing). We derive closed form expressions for each vertices; thus, completely characterizing the curve.

Specifically, we show that each point on the distortion-transmission function (i.e. the optimal distortion for a given value of α) is achieved by a communication strategy that is of the following form:

- Let Z_t be the most recently transmitted symbol up to time t . Then, the optimal estimation strategy is

$$g^*(Y_{0:t}) = Z_t.$$

- Let $E_t = X_t - Z_{t-1}$ and $f^{(k)}$ be a threshold-based strategy given by

$$f^{(k)}(X_t, Y_{0:t-1}) = \begin{cases} 1, & \text{if } |E_t| \geq k; \\ 0, & \text{if } |E_t| < k. \end{cases}$$

¹A symmetric Markov chain defined over \mathbb{Z} does not have a stationary distribution. Therefore, in the limit of no transmission, the expected distribution diverges to ∞ .

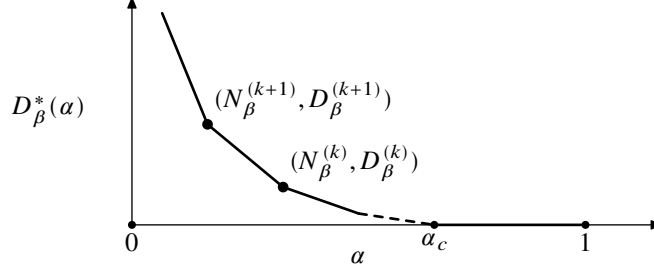


Fig. 3: The distortion-transmission function $D_\beta^*(\alpha)$ for a symmetric Markov source and even and increasing distortion function. $D_\beta^*(\alpha)$ is piecewise linear, convex, and decreasing.

Then, the optimal transmission strategy is a possibly randomized strategy that, at each stage, picks $f^{(k^*)}$ with probability θ^* and picks $f^{(k^*+1)}$ with probability $(1 - \theta^*)$; where k^* is the largest k such that $N_\beta(f^{(k)}, g^*) \geq \alpha$ and θ^* is chosen such that

$$\theta^* N_\beta(f^{(k^*)}, g^*) + (1 - \theta^*) N_\beta(f^{(k^*+1)}, g^*) = \alpha.$$

Note that $f^{(k^*)}(e)$ and $f^{(k^*+1)}(e)$ differ only at $|e| = k^*$. At all other states, they prescribe the same action. Therefore, we can also write f^* as follows:

$$f^*(e) = \begin{cases} 0, & \text{if } |e| < k^*; \\ 0, & \text{w.p. } 1 - \theta^*, \text{ if } |e| = k^*; \\ 1, & \text{w.p. } \theta^*, \text{ if } |e| = k^*; \\ 1, & \text{if } |e| > k^*. \end{cases}$$

The corresponding distortion-transmission function is a piecewise-linear function with vertices given by $(N_\beta^{(k)}, D_\beta^{(k)})$, where

$$D_\beta^{(k)} = D_\beta(f^{(k)}, g^*) \quad \text{and} \quad N_\beta^{(k)} = N_\beta(f^{(k)}, g^*).$$

In addition, $D_\beta^{(1)} = 0$. Therefore,

$$D_\beta^*(\alpha) = 0, \quad \forall \alpha > \alpha_c := N_\beta^{(1)} = \beta(1 - p_0).$$

We show that $\{N_\beta^{(k)}\}_{k=0}^\infty$ is a decreasing sequence and $\{D_\beta^{(k)}\}_{k=0}^\infty$ is an increasing sequence. Consequently, the distortion-transmission function is convex and decreasing. See Fig. 3 for an illustration.

III. PROOF OF THE MAIN RESULT

We proceed as follows. In Sec. III-A, we investigate the Lagrange relaxation of Problems (DIS) and (AVG). Using tools from decentralized stochastic control, in Sec. III-B, we identify the structure of the optimal transmitter and the receiver. In particular, we show that the optimal transmission strategy is of a threshold-type, and the optimal estimation strategy is Kalman-like, and *does not depend on the exact transmission strategy, as long as it is of a threshold-type*. In Sec. III-C, we identify the dynamic programs for the Lagrange relaxations of Problems (DIS) and (AVG). In Sec. III-D and III-E, we provide analytic solutions of these dynamic programs. In particular, we show that the optimal performance is continuous, piecewise linear, concave, and increasing function of the Lagrange multiplier. Using this property, in Sec. III-F, we show that simple Bernoulli randomized strategies (i.e., strategy in which the transmitter randomizes between two actions only in one state) are optimal for the constrained optimization problem. Using this property, we characterize the trade-off between distortion and the number of transmissions.

A. Lagrange relaxations

Problems (DIS) and (AVG) are constrained optimization problems. We first investigate their Lagrange relaxations. For any Lagrange multiplier $\lambda \geq 0$ and any (history dependent) communication strategy (f, g) , define $C_\beta(f, g; \lambda)$ as

$$(1 - \beta) \mathbb{E}^{(f, g)} \left[\sum_{t=0}^{\infty} \beta^t [d(X_t - \hat{X}_t) + \lambda U_t] \mid X_0 = 0 \right]$$

for $\beta \in (0, 1)$, and as

$$\limsup_{T \rightarrow \infty} \mathbb{E}^{(f, g)} \frac{1}{T} \left[\sum_{t=0}^{T-1} [d(X_t - \hat{X}_t) + \lambda U_t] \mid X_0 = 0 \right]$$

for $\beta = 1$.

The Lagrange relaxation of Problems (DIS) and (AVG) is the following: for any $\beta \in (0, 1]$ and $\lambda \geq 0$, find a strategy (f^*, g^*) such that

$$C_\beta^*(\lambda) := C_\beta(f^*, g^*; \lambda) := \inf_{(f, g)} C_\beta(f, g; \lambda) \quad (\text{LAG})$$

where the infimum is taken over all history-dependent communication strategies of the form (1) and (2).

Problem (LAG) is an unconstrained optimization problem with two decision makers—the transmitter and the receiver—that cooperate to minimize a common objective. Such problems are called dynamic team problems or decentralized stochastic control problems [28], [29]. The key challenge in such problems is to identify an appropriate information state or sufficient statistic at each decision maker. Such an information state is then used to identify the structure of optimal communication strategies and a dynamic programming decomposition.

B. Finite horizon setup and the structure of optimal strategies

To identify the structure of the optimal communication strategy, consider the finite-horizon setup of Problem (LAG). Given a time horizon $T \in \mathbb{Z}_{>0}$, a Lagrange multiplier λ , the performance of a strategy (f, g) , where $f = (f_0, \dots, f_T)$ and $g = (g_0, \dots, g_T)$, is given by

$$C_T(f, g; \lambda) := \mathbb{E}^{(f, g)} \left[\sum_{t=0}^T [d(X_t - \hat{X}_t) + U_t] \mid X_0 = 0 \right].$$

The finite-horizon optimization problem is the following: for any $T \in \mathbb{Z}_{>0}$ and $\lambda \geq 0$, find a finite-horizon strategy (f^*, g^*) such that

$$C_T^*(\lambda) := C_T(f^*, g^*; \lambda) = \inf_{(f, g)} C_T(f, g; \lambda) \quad (\text{FIN})$$

where the infimum is taken over all history-dependent communication strategies of the form (1) and (2).

A variation of Problem (FIN) was investigated in [14] (which, in turn, was a variation of [13]) under slightly stronger assumptions:

- (A1') The transition matrix P of the Markov source is a *banded* Toeplitz matrix with decaying off-diagonal terms, i.e., $P_{ij} = p_{|i-j|}$ for $|i-j| \leq b$ and $P_{ij} = 0$ for $|i-j| > b$ for some $b \in \mathbb{Z}_{>0}$; moreover $\{p_1, \dots, p_b\}$ is a decreasing non-negative sequence.
- (A2') The distortion function is either Hamming distortion or k -th mean error, i.e., $d(x - \hat{x})$ is either $\mathbb{1}\{x \neq \hat{x}\}$ or $|x - \hat{x}|^k$.

Under these assumptions, [14] identified the structure of optimal transmission and estimation strategies. Although these results were stated under (A1') and (A2'), the proofs in [14] do not use the specific form of the distortion function but only use the fact that the distortion function is even and increasing on $\mathbb{Z}_{\geq 0}$, i.e. $d(\cdot)$ satisfies (A2). However, assumption (A1') is more critical. Assumption (A1') and the finiteness of time horizon T implies that the reachable set of the state of the Markov sources lies within a finite interval of \mathbb{Z} . This finiteness of the reachable state space was critical to derive the results of [14].

We are interested in infinite horizon setups. As a first step, we generalize the results of [14] by removing the finite support (or banded) assumption in (A1') and show that the same structure is also optimal under the slightly more general assumptions (A1) and (A2). To state these result, we define the following process.

Definition 1 Let Z_t denote the most recently transmitted value of the Markov source. The process $\{Z_t\}_{t=0}^\infty$ evolves in a controlled Markov manner as follows:

$$Z_0 = 0,$$

and

$$Z_t = \begin{cases} X_t, & \text{if } U_t = 1; \\ Z_{t-1}, & \text{if } U_t = 0. \end{cases}$$

□

Note that since U_t can be inferred from the transmitted symbol Y_t , the receiver can also keep track of Z_t as follows:

$$Z_0 = 0,$$

and

$$Z_t = \begin{cases} Y_t, & \text{if } Y_t \neq \mathfrak{E}; \\ Z_{t-1}, & \text{if } Y_t = \mathfrak{E}. \end{cases}$$

Theorem 1 Consider Problem (FIN) under assumptions (A1) and (A2). The process $\{Z_t\}_{t=0}^T$ is a sufficient statistic at the receiver and an optimal estimation strategy is given by

$$\hat{X}_t = g_t^*(Z_t) = Z_t. \quad (3)$$

In general, the optimal estimation strategy depends on the choice of the transmission strategy and vice-versa. Theorem 1 shows that when the Markov process and the distortion function satisfy appropriate symmetry assumptions, the optimal estimation strategy can be specified in closed form. Consequently, we can fix the receiver to be of the above form, and consider the centralized problem of identifying the best transmission strategy.

Definition 2 Let $E_t = X_t - Z_{t-1}$. The process $\{E_t\}_{t=0}^\infty$ evolves in a controlled Markov manner as follows:

$$E_0 = 0,$$

and

$$\mathbb{P}(E_{t+1} = n \mid E_t = e, U_t = u) = \begin{cases} P_{0n}, & \text{if } u = 1; \\ P_{en}, & \text{if } u = 0. \end{cases}$$

□

Theorem 2 Consider Problem (FIN) under assumptions (A1) and (A2) and an estimation strategy given by (3). The process $\{E_t\}_{t=0}^T$ is a sufficient statistic at the transmitter and an optimal transmission strategy is characterized by a sequence of thresholds $\{k_t\}_{t=0}^T$, i.e.,

$$U_t = f_t(E_t) = \begin{cases} 1, & \text{if } |E_t| \geq k_t; \\ 0, & \text{if } |E_t| < k_t. \end{cases} \quad (4)$$

Such an optimal strategy is given by the solution of the following dynamic program:

$$V_{T+1}(\cdot) = 0; \quad (5)$$

and for $t = T, \dots, 0$ and $e \in \mathbb{Z}$,

$$V_t(e; \lambda) = \min \left\{ \lambda + \sum_{n=-\infty}^{\infty} P_{0n} V_{t+1}(n; \lambda), \quad d(e) + \sum_{n=-\infty}^{\infty} P_{en} V_{t+1}(n; \lambda) \right\} \quad (6)$$

where the first term corresponds to choosing $U_t = 1$ and the second term corresponds to choosing $U_t = 0$. Furthermore,

$$C_T^*(\lambda) = V_0(0; \lambda).$$

□

The above structural results were obtained in [14, Theorems 2 and 3] under assumptions (A1') and (A2). We generalize the key steps of the proof presented in [14] to assumptions (A1) and (A2) in Appendix A. Similar results were obtained for Problem (FIN) for Gauss-Markov and autoregressive sources in [13], [15].

Proposition 1 For every $t \in \mathbb{Z}_{>0}$ and $\lambda \geq 0$, the value function $V_t(\cdot; \lambda)$ defined in Theorem 2 is even and increasing on $\mathbb{Z}_{\geq 0}$. □

This is proved in Appendix B.

C. Dynamic programs for the Lagrange relaxations

The structural results of Theorems 1 and 2 extend to the infinite horizon setup as well. The optimal estimation strategy is completely specified by Theorem 1. Note that the optimal estimation strategy does not depend on the choice of the transmission strategy. Therefore, we can fix the estimation strategy and find the transmission strategy that is the *best response* to this estimation strategy. Identifying such a best response strategy is a centralized stochastic control problem. Since the optimal estimation strategy is time-homogeneous, one expects the optimal transmission strategy (i.e., the choice of the optimal thresholds $\{k_t\}_{t=0}^\infty$) to be time-homogeneous as well. To establish such a result, we need the following technical assumption.

(A3) For every $\lambda \geq 0$, there exists a function $w : \mathbb{Z} \rightarrow \mathbb{R}$ and positive and finite constants μ_1 and μ_2 such that for all $e \in \mathbb{Z}$, we have that

$$\max\{\lambda, d(e)\} \leq \mu_1 w(e),$$

and

$$\max \left\{ \sum_{n=-\infty}^{\infty} P_{en} w(n), \sum_{n=-\infty}^{\infty} P_{0n} w(n) \right\} \leq \mu_2 w(e).$$

Remark 1 The model of Example 1 satisfies (A3) with $w(e) = \max\{\lambda, |e|\}$, $\mu_1 = 1$, and $\mu_2 = \max\{1 - 2p + 2p/\lambda, 2\}$. This may be verified by direct substitution. □

Theorem 3 Consider Problem (LAG) for $\beta \in (0, 1)$ under assumptions (A1), (A2), and (A3) and an estimation strategy given by (3). The process $\{E_t\}_{t=0}^\infty$ is a sufficient statistic at the transmitter and an optimal transmission strategy is characterized by a time-homogeneous threshold k , i.e.,

$$U_t = f(E_t) = \begin{cases} 1, & \text{if } |E_t| \geq k; \\ 0, & \text{if } |E_t| < k. \end{cases} \quad (7)$$

Moreover, such an optimal strategy is determined by the unique fixed point of the following dynamic program:

$$V_\beta(e; \lambda) = \min \left\{ (1 - \beta)\lambda + \beta \sum_{n=-\infty}^{\infty} P_{0n} V_\beta(n; \lambda), \quad (1 - \beta)d(e) + \beta \sum_{n=-\infty}^{\infty} P_{en} V_\beta(n; \lambda) \right\}. \quad (8)$$

Let $f_\beta^*(e; \lambda)$ denote the arg min of the right hand side of the above equation. Then the time-homogeneous transmission strategy $f_\beta = \{f_\beta^*(\cdot; \lambda), f_\beta^*(\cdot; \lambda), \dots\}$ is optimal for Problem (LAG) and the given choice of λ and $\beta \in (0, 1)$. Furthermore,

$$C_\beta^*(\lambda) = V_\beta(0; \lambda). \quad \square$$

Proof: The result is the natural extensions of the result of Theorem 2 to the infinite horizon discounted cost setup. The result follows from [30, Proposition 6.10.3]. Note that Assumption (A3) is equivalent to [30, Assumptions 6.10.1, 6.10.2] used in [30, Proposition 6.10.3]. ■

The fixed point $V_\beta(e; \lambda)$ may be computed using value iteration. Since monotonicity is preserved under limits, an immediate consequence of Proposition 1 is the following.

Proposition 2 The value function $V_\beta(\cdot; \lambda)$ defined in Theorem 3 is even and increasing on $\mathbb{Z}_{\geq 0}$. □

We use the vanishing discount approach to extend the results of Theorem 2 to long-term average cost setup; that is, we show that an optimal strategy for the long-term average cost setup may be determined as a limit of the optimal strategy of the discounted cost setup as the discount factor $\beta \uparrow 1$. To use this approach, we show that the value function satisfies the so called SEN conditions of [31].

Proposition 3 For any $\lambda \geq 0$, the value function $V_\beta(\cdot; \lambda)$ satisfies the SEN conditions:

- (S1) There exists a reference state $e_0 \in \mathbb{Z}$ such that $V_\beta(e_0, \lambda) < \infty$ for all $\beta \in (0, 1)$.
- (S2) Define $h_\beta(e; \lambda) = (1 - \beta)^{-1}[V_\beta(e; \lambda) - V_\beta(e_0; \lambda)]$. There exists a function $K_\lambda : \mathbb{Z} \rightarrow \mathbb{R}$ such that $h_\beta(e; \lambda) \leq K_\lambda(e)$ for all $e \in \mathbb{Z}$ and $\beta \in (0, 1)$.
- (S3) There exists a non-negative (finite) constant L_λ such that $-L_\lambda \leq h_\beta(e; \lambda)$ for all $e \in \mathbb{Z}$ and $\beta \in (0, 1)$. □

We prove the result for reference state $e_0 = 0$ in Appendix C.

Theorem 4 Consider Problem (LAG) for $\beta = 1$ under assumptions (A1), (A2), and (A3) and an estimation strategy given by (3). The optimal transmission strategy is time-homogeneous threshold strategy of the form (7). In particular:

- 1) Let $f_1^*(\cdot; \lambda)$ be any limit point of $f_\beta^*(\cdot; \lambda)$ as $\beta \uparrow 1$. Then the time-homogeneous transmission strategy f_1 given as

$$f_1 = \{f_1^*(\cdot; \lambda), f_1^*(\cdot; \lambda), \dots\}$$

is optimal for Problem (LAG) with $\beta = 1$.

- 2) Furthermore, the performance of this optimal strategy is given by

$$C_1^*(\lambda) = \lim_{\beta \uparrow 1} V_\beta(0; \lambda) = \lim_{\beta \uparrow 1} C_\beta^*(\lambda). \quad \square$$

Proof: Since the value function of the discounted cost setup satisfies the SEN conditions, the result follows from [31, Theorem 7.2.3]. In the second part, we use the fact that $e_0 = 0$ was the reference state for the SEN conditions in Proposition 3. ■

Next, we provide analytic solutions of the above dynamic programs. For the discounted cost setup, we start by deriving the performance of an arbitrary time-homogeneous threshold-based strategy of the form (7) and then identify the best strategy in that class. We then use the vanishing discount approach to identify the optimal strategy for the long-term average setup. All of these results are derived under assumptions (A1)–(A3).

D. Analytic solution of the discounted Lagrange relaxation

In this section, we consider the discount factor $\beta \in (0, 1)$. Let \mathcal{F} denote the class of all time-homogeneous threshold-based strategies of the form (7). Let $f^{(k)} \in \mathcal{F}$ denote the strategy with threshold k , $k \in \mathbb{Z}_{\geq 0}$, i.e.,

$$f^{(k)}(e) := \begin{cases} 1, & \text{if } |e| \geq k; \\ 0, & \text{if } |e| < k. \end{cases}$$

Let $D_\beta^{(k)}(e)$ and $N_\beta^{(k)}(e)$ denote the expected discounted distortion and the expected discounted number of transmissions under strategy $f^{(k)}$ when the system starts in state e . Thus,

$$D_\beta^{(k)}(0) = D_\beta(f^{(k)}, g^*), \quad \text{and} \quad N_\beta^{(k)}(0) = N_\beta(f^{(k)}, g^*).$$

From standard results in Markov decision theory, $D_\beta^{(k)}(e)$ and $N_\beta^{(k)}(e)$ are the unique fixed points of the following equations:

$$D_\beta^{(k)}(e) = \begin{cases} \beta \sum_{n=-\infty}^{\infty} P_{0n} D_\beta^{(k)}(n), & \text{if } |e| \geq k; \\ (1 - \beta)d(e) + \beta \sum_{n=-\infty}^{\infty} P_{en} D_\beta^{(k)}(n), & \text{if } |e| < k, \end{cases} \quad (9)$$

and

$$N_\beta^{(k)}(e) = \begin{cases} (1 - \beta) + \beta \sum_{n=-\infty}^{\infty} P_{0n} N_\beta^{(k)}(n), & \text{if } |e| \geq k; \\ \beta \sum_{n=-\infty}^{\infty} P_{en} N_\beta^{(k)}(n), & \text{if } |e| < k. \end{cases} \quad (10)$$

Similarly, let $C_\beta^{(k)}(e; \lambda)$ denote the performance of strategy $f^{(k)}$ for the Lagrange relaxation with discount factor $\beta \in (0, 1)$ and Lagrange multiplier $\lambda \geq 0$ when the system starts in state e . Then,

$$C_\beta^{(k)}(e; \lambda) = D_\beta^{(k)}(e) + \lambda N_\beta^{(k)}(e), \quad (11)$$

and

$$C_\beta^{(k)}(0; \lambda) = C_\beta(f^{(k)}, g^*; \lambda). \quad (12)$$

Let $\tau^{(k)}$ denotes the stopping time when the Markov process with transition probability P starting at state 0 at time $t = 0$ enters the set $\{e \in \mathbb{Z} : |e| \geq k\}$. Note that

$$\tau^{(0)} = 1, \quad \text{and} \quad \tau^{(\infty)} = \infty.$$

Define

$$L_\beta^{(k)} := \mathbb{E} \left[\sum_{t=0}^{\tau^{(k)}-1} \beta^t d(E_t) \mid E_0 = 0 \right] \quad (13)$$

and

$$M_\beta^{(k)} := \frac{1 - \mathbb{E}[\beta^{\tau^{(k)}} \mid E_0 = 0]}{1 - \beta}. \quad (14)$$

We have the following characterization of $D_\beta(f^{(k)}, g^*)$, $N_\beta(f^{(k)}, g^*)$ and $C_\beta(f^{(k)}, g^*; \lambda)$ in terms of $L_\beta^{(k)}$ and $M_\beta^{(k)}$.

Proposition 4 *For any $\beta \in (0, 1)$, the performance of strategy $f^{(k)}$ for the discounted cost Lagrange relaxation is given as follows:*

1) For $k = 0$,

$$D_\beta(f^{(k)}, g^*) = 0, \quad N_\beta(f^{(k)}, g^*) = 1,$$

and

$$C_\beta(f^{(k)}, g^*; \lambda) = \lambda.$$

2) For $k \in \mathbb{Z}_{>0}$

$$D_\beta(f^{(k)}, g^*) = \frac{L_\beta^{(k)}}{M_\beta^{(k)}}, \quad N_\beta(f^{(k)}, g^*) = \frac{1}{M_\beta^{(k)}} - (1 - \beta),$$

and

$$C_\beta(f^{(k)}, g^*; \lambda) = \frac{L_\beta^{(k)} + \lambda}{M_\beta^{(k)}} - \lambda(1 - \beta).$$

□

This is proved in Appendix D.

We can give explicit expressions for $L_\beta^{(k)}$ and $M_\beta^{(k)}$ in terms of the transition matrix and the distortion function. For that matter, define square matrices $P^{(k)}$ and $Q_\beta^{(k)}$ and a column vector $d^{(k)}$ that are indexed by $S^{(k)} := \{-(k-1), \dots, k-1\}$ as follows:

$$P_{ij}^{(k)} := P_{ij}, \quad \forall i, j \in S^{(k)}, \quad (15)$$

$$Q_\beta^{(k)} := [I_{2k-1} - \beta P^{(k)}]^{-1}, \quad (16)$$

$$d^{(k)} := [d(-k+1), \dots, d(k-1)]^\top. \quad (17)$$

Since $\max_i \sum_{j \in S^{(k)}} \beta |P_{ij}^{(k)}| < 1$, $\beta P^{(k)}$ is a *transient* sub-stochastic matrix and by [32, Lemma 1.2.1], $[I_{2k-1} - \beta P^{(k)}]^{-1}$ exists.

Proposition 5 For any $\beta \in (0, 1)$, $L_\beta^{(k)}$ and $M_\beta^{(k)}$ are given by

$$L_\beta^{(k)} = \langle [Q_\beta^{(k)}]_0, d^{(k)} \rangle, \quad (18)$$

$$M_\beta^{(k)} = \langle [Q_\beta^{(k)}]_0, \mathbf{1}_{2k-1} \rangle \quad (19)$$

where $[Q_\beta^{(k)}]_0$ denotes the row with index 0 in $Q_\beta^{(k)}$. Furthermore,

$$L_\beta^{(k)} < L_\beta^{(k+1)}, \quad M_\beta^{(k)} < M_\beta^{(k+1)} \quad (20)$$

and

$$D_\beta^{(k)}(e) < D_\beta^{(k+1)}(e), \quad \forall e \in \mathbb{Z}. \quad (21)$$

This is proved in Appendix E.

Substituting the expressions for $L_\beta^{(k)}$ and $M_\beta^{(k)}$ from Proposition 5 in Proposition 4 gives an explicit analytic expressions for $D_\beta(f^{(k)}, g^*)$ and $N_\beta(f^{(k)}, g^*)$.

An immediate consequence of the above expressions is the following:

Corollary 1 For any $\beta \in (0, 1)$, Let $p_0 = P_{00}$. Then,

$$D_\beta(f^{(1)}, g^*) = 0 \quad \text{and} \quad N_\beta(f^{(1)}, g^*) = \beta(1 - p_0). \quad \square$$

Next, we characterize the optimal strategy using an approach that is inspired by the idea of *calibration* in multi-armed bandits [33]. Let $\lambda_\beta^{(k)}$ be the value of the Lagrange multiplier for which one is indifferent between strategies $f^{(k)}$ and $f^{(k+1)}$ when starting from state 0, i.e., $\lambda_\beta^{(k)}$ is such that

$$C_\beta^{(k)}(0; \lambda_\beta^{(k)}) = C_\beta^{(k+1)}(0; \lambda_\beta^{(k)}). \quad (22)$$

Such a sequence of $\{\lambda_\beta^{(k)}\}_{k=0}^\infty$ can be computed based on $D_\beta^{(k)}(0)$ and $N_\beta^{(k)}(0)$ as follows:

Proposition 6 For any $\beta \in (0, 1)$, the sequence $\{\lambda_\beta^{(k)}\}_{k=0}^\infty$ is given by

$$\lambda_\beta^{(k)} := \frac{D_\beta^{(k+1)}(0) - D_\beta^{(k)}(0)}{N_\beta^{(k)}(0) - N_\beta^{(k+1)}(0)} \quad (23)$$

and satisfies (22) for all $k \in \mathbb{Z}_{\geq 0}$. Under (A2), $\lambda_\beta^{(k)} > 0$ for all $k \in \mathbb{Z}_{\geq 0}$. □

Proof: The expression for $\lambda_\beta^{(k)}$ may be obtained by substituting the result of Proposition 4 in (22). The numerator and the denominator are positive due to Proposition 5. Hence, $\lambda_\beta^{(k)}$ is positive. ■

The optimal strategy is characterized under the following assumption.

(A4) The sequence $\{\lambda_\beta^{(k)}\}_{k=0}^\infty$ defined in Proposition 6 is increasing.

Theorem 5 Consider Problem (LAG) for $\beta \in (0, 1)$ under assumptions (A1)–(A4).

- 1) For all $\lambda \in (\lambda_\beta^{(k)}, \lambda_\beta^{(k+1)})$ such that $\lambda_\beta^{(k)} \neq \lambda_\beta^{(k+1)}$, the strategy $f^{(k+1)}$ is discounted cost optimal.
- 2) The optimal Lagrange performance $C_\beta^*(\lambda)$ is piecewise linear, continuous, concave, and increasing function of λ . □

This is proved in Appendix F. The results of Proposition 6 and Theorem 5 are illustrated in Fig. 4.

E. Analytic solution of the long-term average Lagrange relaxation

As in the discounted cost setup, to find the optimal strategy, we first characterize the performance of a generic threshold-based strategy $f^{(k)} \in \mathcal{F}$. Define for $k \in \mathbb{Z}_{\geq 0}$

$$Q_1^{(k)} := \lim_{\beta \uparrow 1} Q_\beta^{(k)} = [I_{2k-1} - P^{(k)}]^{-1}, \quad (24)$$

$$L_1^{(k)} := \lim_{\beta \uparrow 1} L_\beta^{(k)} = \langle [Q_1^{(k)}]_0, d^{(k)} \rangle, \quad (25)$$

$$M_1^{(k)} := \lim_{\beta \uparrow 1} M_\beta^{(k)} = \langle [Q_1^{(k)}]_0, \mathbf{1}_{2k-1} \rangle. \quad (26)$$

As before, $Q_1^{(k)}$ exists because $P^{(k)}$ is a transient sub-stochastic matrix, i.e., $[I_{2k-1} - P^{(k)}]$ is non-singular.

As in the discounted setup, let $D_1^{(k)}$ and $N_1^{(k)}$ denote the long-term average distortion and the long-term average number of transmissions under strategy $f^{(k)}$ when the system starts in state 0. Similarly, let $C_1^{(k)}(\lambda)$ denote the performance of strategy $f^{(k)}$ for the Lagrange relaxation for the long-term average setup, starting at initial state 0.

Proposition 7 *The performance of strategy $f^{(k)}$ for the long-term average cost Lagrange relaxation is given by*

$$\begin{aligned} D_1^{(k)} &= \lim_{\beta \uparrow 1} D_\beta^{(k)}(0) = \frac{L_1^{(k)}}{M_1^{(k)}}; \\ N_1^{(k)} &= \lim_{\beta \uparrow 1} N_\beta^{(k)}(0) = \frac{1}{M_1^{(k)}}; \\ C_1^{(k)}(\lambda) &= \lim_{\beta \uparrow 1} C_\beta^{(k)}(0; \lambda) = \frac{L_1^{(k)} + \lambda}{M_1^{(k)}}. \end{aligned} \quad \square$$

Proof: This result can be proved by an argument similar to Theorem 4. Similar to Proposition 3, we can show that $D_\beta^{(k)}(e)$ and $N_\beta^{(k)}(e)$ satisfy the SEN conditions. Then, the result follows from [31, Theorem 7.2.3]. ■

As for the discounted case, we can show that

Proposition 8 $L_1^{(k)} < L_1^{(k+1)}$ and $M_1^{(k)} < M_1^{(k+1)}$. □

Note that if we simply use the result of Proposition 5 and take limit over β , we will not get the strict inequality given in Proposition 8. Nonetheless, the strict inequality follows by an argument similar to that in the proof of Proposition 5.

Similar to Corollary 1, we have the following:

Corollary 2 *Let $p_0 = P_{00}$. Then,*

$$D_1(f^{(1)}, g^*) = 0 \quad \text{and} \quad N_1(f^{(1)}, g^*) = (1 - p_0). \quad \square$$

To characterize the optimal strategy, define

$$\lambda_1^{(k)} := \lim_{\beta \uparrow 1} \lambda_\beta^{(k)} = \frac{D_1^{(k+1)} - D_1^{(k)}}{N_1^{(k)} - N_1^{(k+1)}}. \quad (27)$$

Note that (A4) implies that $\{\lambda_1^{(k)}\}_{k=0}^\infty$ is increasing. Then, we have the following:

Theorem 6 . *Consider Problem (LAG) for $\beta = 1$ under assumptions (A1)–(A4).*

- 1) *For all $\lambda \in (\lambda_1^{(k)}, \lambda_1^{(k+1)}]$ such that $\lambda_1^{(k)} \neq \lambda_1^{(k+1)}$, the strategy $f^{(k+1)}$ is long-term average cost optimal.*
- 2) *The optimal Lagrange performance $C_1^*(\lambda)$ is continuous, piecewise linear, concave, and increasing function of λ .* □

This is proved in Appendix G. Also see Fig. 4.

F. The constrained optimization problems

Finally, we come back to the constrained optimization problems (DIS) and (AVG). To describe the solution of these problems, we first define Bernoulli randomized strategy and Bernoulli randomized simple strategy.

Definition 3 Suppose we are given two (non-randomized) time-homogeneous strategies f_1 and f_2 and a randomization parameter $\theta \in (0, 1)$. The Bernoulli randomized strategy (f_1, f_2, θ) is a strategy that randomizes between f_1 and f_2 at each stage; choosing f_1 with probability θ and f_2 with probability $(1 - \theta)$. Such a strategy is called a Bernoulli randomized simple strategy if f_1 and f_2 differ on exactly one state i.e. there exists a state e_0 such that

$$f_1(e) = f_2(e), \quad \forall e \neq e_0. \quad \square$$

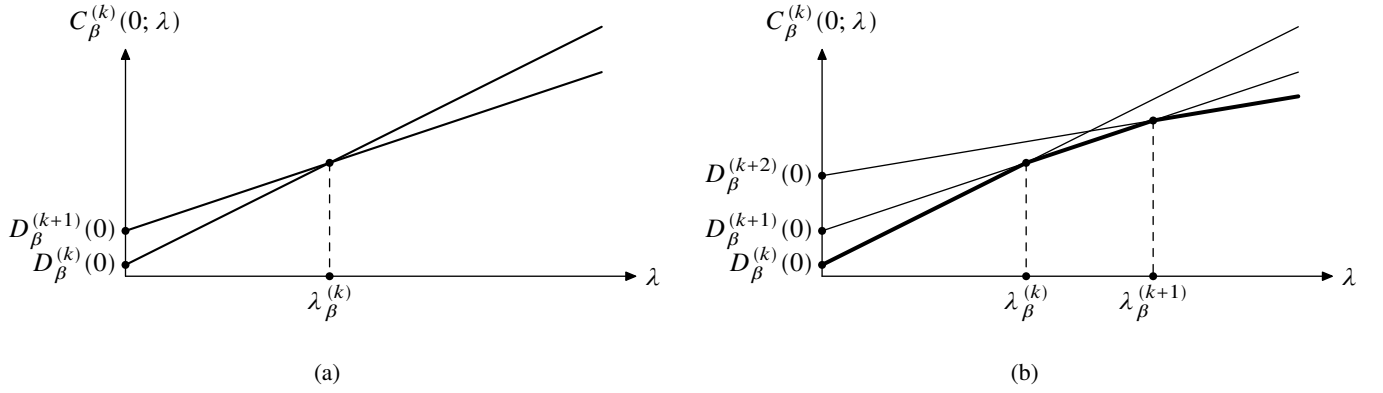


Fig. 4: Plot (a) shows $C_\beta^{(k)}(0; \lambda)$ and $C_\beta^{(k+1)}(0; \lambda)$. $\lambda_\beta^{(k)}$ is the x -coordinate of the intersection of these two lines. Plot (b) shows $C_\beta^*(\lambda)$ in bold. As is evident from the plot, $C_\beta^*(\lambda)$ is piecewise linear, concave and increasing. Moreover, note that $C_\beta^{(k+1)}(0; \lambda)$ is the smallest among $\{C_\beta^{(k)}(0; \lambda)\}_{k=0}^\infty$ when $\lambda \in (\lambda_\beta^{(k)}, \lambda_\beta^{(k+1)})$. Hence, $f^{(k+1)}$ is optimal for that range of λ .

Define

$$k_\beta^*(\alpha) = \sup\{k \in \mathbb{Z}_{\geq 0} : N_\beta(f^{(k)}, g^*) \geq \alpha\} \quad (28)$$

and

$$\theta_\beta^*(\alpha) = \frac{\alpha - N_\beta(f^{(k_\beta^*(\alpha)+1)}, g^*)}{N_\beta(f^{(k_\beta^*(\alpha))}, g^*) - N_\beta(f^{(k_\beta^*(\alpha)+1)}, g^*)}. \quad (29)$$

For ease of notation, we use $k^* = k_\beta^*(\alpha)$ and $\theta^* = \theta_\beta^*(\alpha)$. By definition, $\theta^* \in [0, 1]$ and

$$\theta^* N_\beta(f^{(k^*)}, g^*) + (1 - \theta^*) N_\beta(f^{(k^*+1)}, g^*) = \alpha. \quad (30)$$

Note that k^* and θ^* could have been equivalently defined as follow:

$$k^* = \sup\left\{k \in \mathbb{Z}_{\geq 0} : M_\beta^{(k)} \leq \frac{1}{1 + \alpha - \beta}\right\}, \quad \theta^* = \frac{M^{(k^*+1)} - \frac{1}{1 + \alpha - \beta}}{M^{(k^*+1)} - M^{(k^*)}}.$$

Theorem 7 Let f^* be the Bernoulli randomized simple strategy $(f^{(k^*)}, f^{(k^*+1)}, \theta^*)$. i.e.

$$f^*(e) = \begin{cases} 0, & \text{if } |e| < k^*; \\ 0, & \text{w.p. } 1 - \theta^*, \text{ if } |e| = k^*; \\ 1, & \text{w.p. } \theta^*, \text{ if } |e| = k^*; \\ 1, & \text{if } |e| > k^*. \end{cases} \quad (31)$$

Then (f^*, g^*) is optimal for the constrained Problem (DIS) when $\beta \in (0, 1)$ and Problem (AVG) when $\beta = 1$. \square

Proof: The proof relies on the following characterization of the optimal strategy stated in [34, Proposition 1.2]. The characterization was stated for the long-term average setup but a similar result can be shown for the discounted case as well, for example, by using the approach of [35]. Also, see [36, Theorem 8.1] for a similar sufficient condition for general constrained optimization problem.

A (possibly randomized) strategy (f°, g°) is optimal for a constrained optimization problem with $\beta \in (0, 1]$ if the following conditions hold:

(C1) $N_\beta(f^\circ, g^\circ) = \alpha$,

(C2) There exists a Lagrange multiplier $\lambda^\circ \geq 0$ such that (f°, g°) is optimal for $C_\beta(f, g; \lambda^\circ)$.

We will show that the strategies (f^*, g^*) satisfy (C1) and (C2) with $\lambda^\circ = \lambda_\beta^{(k^*)}$.

(f^*, g^*) satisfy (C1) due to (30). For $\lambda = \lambda_\beta^{(k^*)}$, both $f^{(k^*)}$ and $f^{(k^*+1)}$ are optimal for $C_\beta(f, g; \lambda)$. Hence, any strategy randomizing between them, in particular f^* , is also optimal for $C_\beta(f, g; \lambda)$. Hence (f^*, g^*) satisfies (C2). Therefore, by [34, Proposition 1.2], (f^*, g^*) is optimal for Problems (DIS) and (AVG). \blacksquare

Theorem 8 The distortion-transmission function is given by

$$D_\beta^*(\alpha) = \theta^* D_\beta(f^{(k^*)}, g^*) + (1 - \theta^*) D_\beta(f^{(k^*+1)}, g^*). \quad (32)$$

Furthermore, $D_\beta^*(\alpha)$ is a continuous, piecewise linear, decreasing, and convex function of α . \square

Proof: The form of $D_\beta^*(\alpha)$ given in (32) follows immediately from the fact that (f^*, g^*) is a Bernoulli randomized simple strategy. As argued in Section III-F, $D_\beta^*(\alpha)$ will always be decreasing and convex in α .

For any $k \in \mathbb{Z}_{\geq 0}$, define

$$\alpha^{(k)} = N_\beta(f^{(k)}, g^*),$$

and consider any $\alpha \in (\alpha^{(k+1)}, \alpha^{(k)})$. Then,

$$k_\beta^*(\alpha^{(k)}) = k, \quad \text{and} \quad \theta_\beta^*(\alpha^{(k)}) = 1.$$

Hence

$$D_\beta^*(\alpha^{(k)}) = D_\beta(f^{(k)}, g^*).$$

Thus, by (29)

$$\theta^* = \frac{\alpha - \alpha^{(k+1)}}{\alpha^{(k)} - \alpha^{(k+1)}},$$

and by (32),

$$D_\beta^*(\alpha) = \theta^* D_\beta^*(\alpha^{(k)}) + (1 - \theta^*) D_\beta^*(\alpha^{(k+1)}).$$

Therefore $D_\beta^*(\alpha)$ is piecewise linear and continuous. \blacksquare

It follows from the argument given in the proof above that $\{(\alpha^{(k)}, D_\beta^*(\alpha^{(k)}))\}_{k=0}^\infty$ are the vertices of the piecewise linear function D_β^* . See Fig. 3 for an illustration.

Combining Theorem 7 with the results of Corollaries 1 and 2, we get

Corollary 3 Let $p_0 = P_{00}$. Then,

$$D_\beta^*(\alpha) = 0, \quad \forall \alpha \geq \alpha_c := \beta(1 - p_0). \quad \square$$

IV. AN EXAMPLE: APERIODIC, SYMMETRIC BIRTH-DEATH MARKOV CHAIN

In this section, we characterize $D_\beta^*(\alpha)$ for the birth-death Markov chain presented in Example 1. As shown in Remark 1, this model satisfies Assumption (A3). Thus, we can use Proposition 6 and (23) to compute the critical Lagrange multipliers $\{\lambda_\beta^{(k)}\}_{k=0}^\infty$. The results of Theorems 5 and 6 are given in terms of $L_\beta^{(k)}$ and $M_\beta^{(k)}$, which, in turn, depend on the matrix $Q_\beta^{(k)}$. The matrix $Q_\beta^{(k)}$ is the inverse of a tridiagonal symmetric Toeplitz matrix and an explicit formula for its elements is available [37].

Lemma 1 Define for $\beta \in (0, 1]$

$$K_\beta = -2 - \frac{(1 - \beta)}{\beta p} \quad \text{and} \quad m_\beta = \cosh^{-1}(-K_\beta/2)$$

Then,

$$[Q_\beta^{(k)}]_{ij} = \frac{1}{\beta p} \frac{[A_\beta^{(k)}]_{ij}}{b_\beta^{(k)}}, \quad i, j \in S^{(k)},$$

where, for $\beta \in (0, 1)$,

$$\begin{aligned} [A_\beta^{(k)}]_{ij} &= \cosh((2k - |i - j|)m_\beta) - \cosh((i + j)m_\beta), \\ b_\beta^{(k)} &= \sinh(m_\beta) \sinh(2km_\beta); \end{aligned}$$

and for $\beta = 1$,

$$\begin{aligned} [A_1^{(k)}]_{ij} &= (k - \max\{i, j\})(k + \min\{i, j\}), \\ b_1^{(k)} &= 2k. \end{aligned} \quad \square$$

In particular, the elements $[Q_\beta^{(k)}]_{0j}$ are given as follows. For $\beta \in (0, 1)$,

$$[Q_\beta^{(k)}]_{0j} = \frac{1}{\beta p} \frac{\cosh((2k - |j|)m_\beta) - \cosh(jm_\beta)}{2 \sinh(m_\beta) \sinh(2km_\beta)}, \quad (33)$$

and for $\beta = 1$,

$$[Q_1^{(k)}]_{0j} = \frac{(k - \max\{j, 0\})(k + \min\{j, 0\})}{2pk}. \quad (34)$$

TABLE I: Values of $D_\beta^{(k)}$, $N_\beta^{(k)}$ and $\lambda_\beta^{(k)}$ for different values of k and β for the birth-death Markov chain of Example 1 with $p = 0.3$.

(a) For $\beta = 0.9$				(b) For $\beta = 0.95$				(c) For $\beta = 1.0$			
k	$D_\beta^{(k)}$	$N_\beta^{(k)}$	$\lambda_\beta^{(k)}$	k	$D_\beta^{(k)}$	$N_\beta^{(k)}$	$\lambda_\beta^{(k)}$	k	$D_\beta^{(k)}$	$N_\beta^{(k)}$	$\lambda_\beta^{(k)}$
0	0	1	0	0	0	1	0	0	0	1	0
1	0	0.5400	1.0989	1	0	0.5700	1.1050	1	0	0.6000	1.1111
2	0.4576	0.1236	4.1021	2	0.4790	0.1365	4.3657	2	0.5000	0.1500	4.6667
3	0.7695	0.0475	9.2839	3	0.8282	0.0565	10.6058	3	0.8889	0.0667	12.3810
4	1.0066	0.0220	16.2509	4	1.1218	0.0288	19.9550	4	1.2500	0.0375	25.9259
5	1.1844	0.0111	24.4478	5	1.3715	0.0163	32.0869	5	1.6000	0.0240	46.9697
6	1.3130	0.0058	33.4121	6	1.5811	0.0098	46.4727	6	1.9444	0.0167	77.1795
7	1.4029	0.0031	42.8289	7	1.7536	0.0061	62.5651	7	2.2857	0.0122	118.2222
8	1.4638	0.0017	52.5042	8	1.8927	0.0039	79.8921	8	2.6250	0.0094	171.7647
9	1.5040	0.0009	62.3245	9	2.0028	0.0025	98.0854	9	2.9630	0.0074	239.4737
10	1.5298	0.0005	72.2255	10	2.0884	0.0016	116.8739	10	3.0000	0.0060	323.0159

Proof: The matrix $I_{2k-1} - \beta P^{(k)}$ is a symmetric tridiagonal matrix given by

$$I_{2k-1} - \beta P^{(k)} = -\beta p \begin{bmatrix} K_\beta & 1 & 0 & \cdots & \cdots & 0 \\ 1 & K_\beta & 1 & 0 & \cdots & 0 \\ 0 & 1 & K_\beta & 1 & \cdots & 0 \\ \vdots & \ddots & \ddots & \ddots & \ddots & \vdots \\ 0 & \cdots & 0 & 1 & K_\beta & 1 \\ 0 & 0 & \cdots & 0 & 1 & K_\beta \end{bmatrix}.$$

$Q_\beta^{(k)}$ is the inverse of the above matrix. The inverse of the tridiagonal matrix in the above form with $K_\beta \leq -2$ are computed in closed form in [37]. The result of the lemma follows from these results. ■

Using the expressions for $Q_\beta^{(k)}$, we obtain closed form expressions for $L_\beta^{(k)}$ and $M_\beta^{(k)}$.

Remark 2 For ease of notation, in rest of the paper we write $D_\beta^{(k)}$ and $N_\beta^{(k)}$ in place of $D_\beta^{(k)}(0)$ and $N_\beta^{(k)}(0)$. □

Lemma 2 1) For $\beta \in (0, 1)$,

$$D_\beta^{(k)} = \frac{\sinh(km_\beta) - k \sinh(m_\beta)}{2 \sinh^2(km_\beta/2) \sinh(m_\beta)};$$

$$N_\beta^{(k)} = \frac{2\beta p \sinh^2(m_\beta/2) \cosh(km_\beta)}{\sinh^2(km_\beta/2)} - (1 - \beta).$$

2) For $\beta = 1$,

$$D_1^{(k)} = \frac{k^2 - 1}{3k};$$

$$N_1^{(k)} = \frac{2p}{k^2};$$

and

$$\lambda_1^{(k)} = \frac{k(k+1)(k^2+k+1)}{6p(2k+1)}.$$

□

Proof: By substituting the expression for $Q_\beta^{(k)}$ from Lemma 1 in the expressions for $L_\beta^{(k)}$ and $M_\beta^{(k)}$ from Proposition 5 (and the corresponding expressions for $\beta = 1$), we get that

1) For $\beta \in (0, 1)$,

$$L_\beta^{(k)} = \frac{\sinh(km_\beta) - k \sinh(m_\beta)}{4\beta p \sinh^2(m_\beta/2) \sinh(m_\beta) \cosh(km_\beta)},$$

$$M_\beta^{(k)} = \frac{\sinh^2(km_\beta/2)}{2\beta p \sinh^2(m_\beta/2) \cosh(km_\beta)}.$$

2) For $\beta = 1$,

$$L_1^{(k)} = k(k^2 - 1)/(6p),$$

$$M_1^{(k)} = k^2/(2p).$$

The results of the lemma follow using the above expressions and Proposition 4 and 7. The expression for $\lambda_1^{(k)}$ is obtained by plugging the expressions of $D_1^{(k+1)}$, $D_1^{(k)}$, $N_1^{(k+1)}$, and $N_1^{(k)}$ in (27). ■

When $p = 0.3$, the values of $D_\beta^{(k)}$, $N_\beta^{(k)}$, and $\lambda_\beta^{(k)}$ for different values of k and β are shown in Table I.

For $\beta = 1$, we can use the analytic expression of $\lambda^{(k)}$ to verify that $\{\lambda_\beta^{(k)}\}_{k=0}^\infty$ is increasing. For $\beta \in (0, 1)$, we can numerically verify that $\{\lambda_\beta^{(k)}\}_{k=0}^\infty$ is increasing. Thus, Assumption (A4) is satisfied and we can use the results of Theorems 5 and 6. For $p = 0.3$, the optimal Lagrange performance for different values of β is shown in Fig. 5.

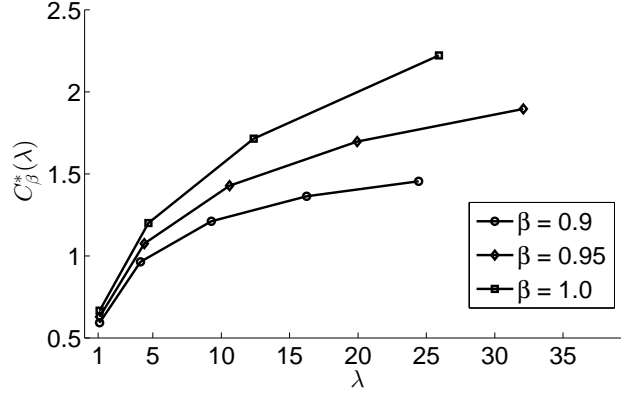


Fig. 5: Plot of $C_\beta^*(\lambda)$ vs λ for the birth-death Markov chain of Example 1 with $p = 0.3$.

Lemma 3 1) For $\beta \in (0, 1)$, k_β^* is given by the maximum k that satisfies the following inequality

$$\frac{2 \cosh(km_\beta)}{\cosh(km_\beta) - 1} \geq \frac{1 + \alpha - \beta}{\beta p (\cosh(m_\beta) - 1)}.$$

2) For $\beta = 1$, k_1^* is given by the following equation

$$k_1^* = \left\lfloor \sqrt{\frac{2p}{\alpha}} \right\rfloor.$$

□

Proof: The result of the lemma follows directly by using the definition of k_β^* given in (28) in the expressions given in Lemma 2. ■

Using the above results, we can plot the distortion-transmission function $D_\beta^*(\alpha)$. See Fig. 6 for the plot of $D_\beta^*(\alpha)$ vs α for different values of β (all for $p = 0.3$). An alternative way to plot this curve is to draw the vertices $(N_\beta^{(k)}, D_\beta^{(k)})$ using the data in Table I to compute the optimal (randomized) strategy for a particular value of α .

As an example, suppose we want to identify the optimal strategy at $\alpha = 0.5$ for the birth-death Markov chain of Example 1 with $p = 0.3$ and $\beta = 0.9$. Recall that k^* is the largest value of k such that $N_\beta^{(k)} \leq \alpha$. Thus, from Table Ia, we get that $k^* = 1$. Then, by (29),

$$\theta^* = \frac{\alpha - N_\beta^{(2)}}{N_\beta^{(1)} - N_\beta^{(2)}} = 0.9039.$$

Let $f^* = (f^{(1)}, f^{(2)}, \theta^*)$. Then the Bernoulli randomized simple strategy (f^*, g^*) is optimal for Problem (DIS). Furthermore, by (32)

$$D_\beta^*(\alpha) = 0.044.$$

V. CONCLUSION

We characterized the distortion-transmission function for transmitting a first-order symmetric Markov source in real-time with constraints on the expected number of transmissions.

Our result depends critically on establishing the following structure of optimal communication strategies.

- (S) There is no loss of optimality in restricting attention to threshold based transmission strategies and as long as the transmission strategy belongs to this class, the optimal estimation strategy is independent of the choice of the threshold.

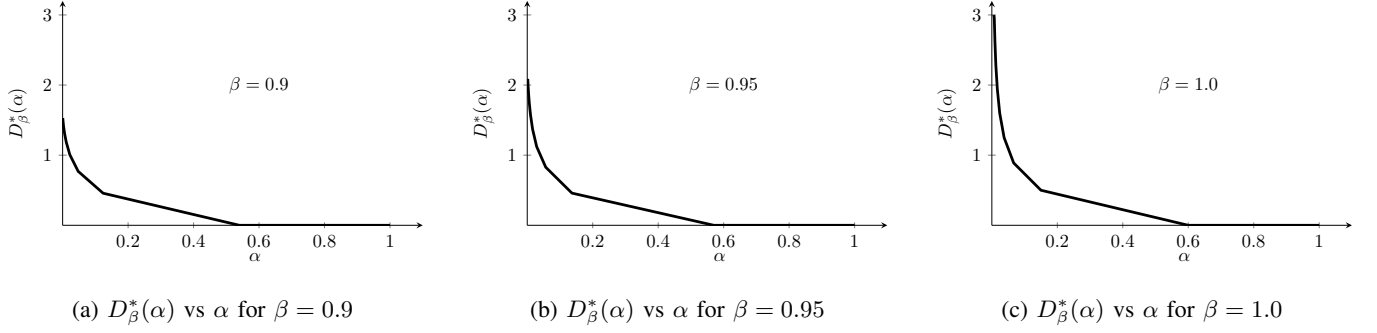


Fig. 6: Plots of $D_\beta^*(\alpha)$ vs α for different β for the birth-death Markov chain of Example 1 with $p = 0.3$.

As a consequence of this structure, the optimal estimation strategy is known, and we only have to identify the optimal transmission strategy. We look at the Lagrange relaxation, compute the performance of an arbitrary threshold based transmission strategy, identify the set of Lagrange multipliers for which an arbitrary threshold based strategy is optimal, and then use these features to identify the optimal strategy for the constrained optimization problem.

A. Salient features of the distortion-transmission function and the optimal strategy

By definition, the distortion-transmission function $D_\beta^*(\alpha)$ is convex and decreasing in the constraint α . We show that, in addition, it has the following features:

- 1) $D_\beta^*(\alpha)$ is piecewise linear in α .
- 2) Any point on $D_\beta^*(\alpha)$ is achieved by a strategy that chooses a randomized action in at most two states.

These features are a consequence of the discreteness of the source. If the source is continuous valued, then $D_\beta^*(\alpha)$ will be smooth and achieved by a pure (non-randomized) strategy.

As an example, consider a scalar Gauss-Markov source. As shown in [13], the structure of optimal transmission and estimation strategies is similar to Theorems 1 and 2. We can follow the approach presented in this paper: consider a threshold strategy $f^{(k)}$ and characterize the distortion $D_\beta^{(k)}$ and the number of transmissions $N_\beta^{(k)}$ under $f^{(k)}$. The main difference will be that since k takes values in \mathbb{R} , instead of (22), $\lambda_\beta^{(k)}$ would be characterized by

$$\lambda_\beta^{(k)} = \arg \min_{\lambda \geq 0} C_\beta^{(k)}(0; \lambda).$$

We would get that under appropriate technical conditions, such a $\lambda_\beta^{(k)}$ exists, and is increasing and continuous in k . In particular, for any α , we can identify a Lagrange multiplier λ° and a threshold k° such that, $N_\beta^{(k^\circ)}(0) = \alpha$ and $f^{(k^\circ)}$ is optimal for $C_\beta^*(f, g; \lambda^\circ)$. Hence, by the argument given in Theorem 7, the pure (non-randomized) strategy $(f^{(k^\circ)}, g^*)$ will be optimal for the constrained optimization problem. In contrast, for the discrete Markov sources, randomization is needed because there may not exist a threshold k° such that $N_\beta^{(k^\circ)}(0) = \alpha$.

B. Comments on the assumptions

The results were derived under the four assumptions (A1)–(A4). Assumption (A1) is a limiting assumption that restricts the results to Markov sources over \mathbb{Z} that satisfy a symmetry property. Assumption (A2) is a mild assumption that restricts the results to even and increasing distortion functions. One expects (A2) to be satisfied in most applications. Assumption (A3) is a mild technical assumption to ensure that the distortion function is not increasing too quickly. Assumption (A4) is a property of the critical Lagrange multiplier that is difficult to verify in general. However, for a specific source and distortion function, like the one presented in Example 1, this assumption can be verified either numerically or analytically. One can also identify sufficient conditions for (A4) (for example, $D_\beta^{(k)}(0)$ is convex in k and $N_\beta^{(k)}(0)$ is concave in k) that might be easier to verify for specific sources.

The critical restrictive assumption is (A1). Some kind of symmetry in the source is needed to use majorization theory to derive the structure (S). One immediate question is whether structure (S) also holds for symmetric sources defined over a finite alphabet (for example, a random walk over $\{1, 2, \dots, n\}$). To obtain such generalizations, we need to define a notion of ASU distributions over a finite alphabet and a notion of majorization that is preserved under additions over that alphabet (see Lemmas 4, 5, and 6). We are not aware of such results.

C. Deterministic implementation

The optimal strategy shown in Theorem 7 chooses a randomized action in states $\{-k^*, k^*\}$. It is also possible to identify deterministic (non-randomized) but time-varying strategies that achieve the same performance. We describe two such strategies for the long-term average setup.

1) *Steering strategies*: Let a_t^0 (respectively, a_t^1) denote the number of times the action $u_t = 0$ (respectively, the action $u_t = 1$) has been chosen in states $\{-k^*, k^*\}$ in the past, i.e.

$$a_t^i = \sum_{s=0}^{t-1} \mathbb{1}\{|E_s| = k^*, u_s = i\}, \quad i \in \{0, 1\}.$$

Thus, the empirical frequency of choosing action $u_t = i$, $i \in \{0, 1\}$, in states $\{-k^*, k^*\}$ is $a_t^i / (a_t^0 + a_t^1)$. A steering strategy compares these empirical frequencies with the desired randomization probabilities $\theta^0 = 1 - \theta^*$ and $\theta^1 = \theta^*$ and chooses an action that *steers* the empirical frequency closer to the desired randomization probability. More formally, at states $\{-k^*, k^*\}$, the steering transmission strategy chooses the action

$$\arg \max_i \left\{ \theta^i - \frac{a_t^i + 1}{a_t^0 + a_t^1 + 1} \right\}$$

in states $\{-k^*, k^*\}$ and chooses deterministic actions according to f^* (given in (31)) in states except $\{-k^*, k^*\}$. Note that the above strategy is deterministic (non-randomized) but depends on the history of visits to states $\{-k^*, k^*\}$. Such strategies were proposed in [38], where it was shown that the steering strategy described above achieves the same performance as the randomized strategy f^* and hence is optimal for Problem (AVG). Variations of such steering strategies have been proposed in [39], [40], where the adaptation was done by comparing the sample path average cost with the expected value (rather than by comparing empirical frequencies).

2) *Time-sharing strategies*: Define a cycle to be the period of time between consecutive visits of process $\{E_t\}_{t=0}^\infty$ to state zero. A time-sharing strategy is defined by a series $\{(a_m, b_m)\}_{m=0}^\infty$ and uses strategy $f^{(k^*)}$ for the first a_0 cycles, uses strategy $f^{(k^*+1)}$ for the next b_0 cycles, and continues to alternate between using strategy $f^{(k^*)}$ for a_m cycles and strategy $f^{(k^*+1)}$ for b_m cycles. In particular, if $(a_m, b_m) = (a, b)$ for all m , then the time-sharing strategy is a periodic strategy that uses $f^{(k^*)}$ a cycles and $f^{(k^*+1)}$ for b cycles.

The performance of such time-sharing strategies was evaluated in [41], where it was shown that if the cycle-lengths of the time-sharing strategy are chosen such that,

$$\begin{aligned} \lim_{M \rightarrow \infty} \frac{\sum_{m=0}^M a_m}{\sum_{m=0}^M (a_m + b_m)} &= \frac{\theta^* N_1^{(k^*)}}{\theta^* N_1^{(k^*)} + (1 - \theta^*) N_1^{(k^*+1)}} \\ &= \frac{\theta^* N_1^{(k^*)}}{\alpha}. \end{aligned}$$

Then the time-sharing strategy $\{(a_m, b_m)\}_{m=0}^\infty$ achieves the same performance as the randomized strategy f^* and hence, is optimal for Problem (AVG).

APPENDIX A PROOF OF THE STRUCTURAL RESULTS

The results of [14] relied on the notion of ASU (almost symmetric and unimodal) distributions introduced in [42].

Definition 4 (Almost symmetric and unimodal distribution) A probability distribution μ on \mathbb{Z} is almost symmetric and unimodal (ASU) about a point $a \in \mathbb{Z}$ if for every $n \in \mathbb{Z}_{\geq 0}$,

$$\mu_{a+n} \geq \mu_{a-n} \geq \mu_{a+n+1}.$$

□

A probability distribution that is ASU around 0 and even (i.e., $\mu_n = \mu_{-n}$) is called ASU and even. Note that the definition of ASU and even is equivalent to even and decreasing on $\mathbb{Z}_{\geq 0}$.

Definition 5 (ASU Rearrangement) The ASU rearrangement of a probability distribution μ , denoted by μ^+ , is a permutation of μ such that for every $n \in \mathbb{Z}_{\geq 0}$,

$$\mu_n^+ \geq \mu_{-n}^+ \geq \mu_{n+1}^+.$$

□

We now introduce the notion of majorization for distributions supported over \mathbb{Z} , as defined in [43].

Definition 6 (Majorization) Let μ and ν be two probability distributions defined over \mathbb{Z} . Then μ is said to majorize ν , which is denoted by $\mu \succeq_m \nu$, if for all $n \in \mathbb{Z}_{\geq 0}$,

$$\begin{aligned} \sum_{i=-n}^n \mu_i^+ &\geq \sum_{i=-n}^n \nu_i^+, \\ \sum_{i=-n}^{n+1} \mu_i^+ &\geq \sum_{i=-n}^{n+1} \nu_i^+. \end{aligned}$$

□

The model considered in [14] was slightly different than the one presented in Section II. Instead of the Markov source with a given transition probability matrix, it was assumed in [14] that the Markov source evolves according to

$$X_{t+1} = X_t + M_t$$

where M_t has an ASU and even distribution with a finite support. Our model is equivalent, except that we assume M_t has an ASU and even distribution with possibly countable support.

The structural results of Theorem 1 were proved in two-steps in [14]. The first step relied on the following two results.

Lemma 4 Let μ and ν be probability distributions with finite support defined over \mathbb{Z} . If μ is ASU and even and ν is ASU about a , then the convolution $\mu * \nu$ is ASU about a . □

Lemma 5 Let μ , ν , and ξ be probability distributions with finite support defined over \mathbb{Z} . If μ is ASU and even, ν is ASU, and ξ is arbitrary, then $\nu \succeq_m \xi$ implies that $\mu * \nu \succeq_m \mu * \xi$. □

These results were originally proved in [42] and were stated as Lemmas 5 and 6 in [14].

The second step (in the proof of Theorem 1) in [14] relied on the following result.

Lemma 6 Let μ be a probability distribution with finite support defined over \mathbb{Z} and $f: \mathbb{Z} \rightarrow \mathbb{R}_{\geq 0}$. Then,

$$\sum_{n=-\infty}^{\infty} f(n) \mu_n \leq \sum_{n=-\infty}^{\infty} f^+(n) \mu_n^+.$$

□

We generalize the results of Lemmas 4, 5, and 6 to distributions over \mathbb{Z} with possibly countable support. With these generalizations, we can follow the same two step approach of [14] to prove Theorem 1.

The proof of Theorem 2 in [14] only relied on the result of Theorem 1. The exact same proof works in our model as well.

A. Generalization of Lemma 4 to distributions supported over \mathbb{Z}

The proof argument is similar to that presented in [42, Lemma 6.2]. We first prove the results for $a = 0$. Assume that ν is ASU and even. For any $n \in \mathbb{Z}_{\geq 0}$, let $r^{(n)}$ denote the rectangular function from $-n$ to n , i.e.,

$$r^{(n)}(e) = \begin{cases} 1, & \text{if } |e| \leq n, \\ 0, & \text{otherwise.} \end{cases}$$

Note that any ASU and even distribution μ may be written as a sum of rectangular functions as follows:

$$\mu = \sum_{n=0}^{\infty} (\mu_n - \mu_{n+1}) r^{(n)}.$$

It should be noted that $\mu_n - \mu_{n+1} \geq 0$ because μ is ASU and even. ν may also be written in a similar form.

The convolution of any two rectangular functions $r^{(n)}$ and $r^{(m)}$ is ASU and even. Therefore, by the distributive property of convolution, the convolution of μ and ν is also ASU and even.

The proof for the general $a \in \mathbb{Z}$ follows from the following facts:

- 1) Shifting a distribution is equivalent to convolution with a shifted delta function.
- 2) Convolution is commutative and associative.

B. Generalization of Lemma 5 to distributions supported over \mathbb{Z}

We follow the proof idea of [43, Theorem II.1]. For any probability distribution μ , we can find distinct indices i_j , $|j| \leq n$ such that $\mu(i_j)$, $|j| \leq n$, are the $2n + 1$ largest values of μ . Define

$$\mu_n(i_j) = \mu(i_j),$$

for $|j| \leq n$ and 0 otherwise. Clearly, $\mu_n \uparrow \mu$ and if μ is ASU and even, so is μ_n .

Now consider the distributions μ , ν , and ξ from Lemma 5 but without the restriction that they have finite support. For every $n \in \mathbb{Z}_{\geq 0}$, define μ_n , ν_n , and ξ_n as above. Note that all distributions have finite support and μ_n is ASU and even and ν_n is ASU. Furthermore, since the definition of majorization remain unaffected by truncation described above, $\nu_n \succeq_m \xi_n$. Therefore, by Lemma 5,

$$\mu_n * \nu_n \succeq_m \mu_n * \xi_n.$$

By taking limit over n and using the monotone convergence theorem, we get

$$\mu * \nu \succeq_m \mu * \xi.$$

C. Generalization of Lemma 6 to distributions supported over \mathbb{Z}

This is an immediate consequence of [43, Theorem II.1].

APPENDIX B PROOF OF PROPOSITION 1

Definition 7 (Stochastic Dominance) Let μ and ν be two probability distributions defined over $\mathbb{Z}_{\geq 0}$. Then μ is said to dominate ν in the sense of stochastic dominance, which is denoted by $\mu \succeq_s \nu$, if

$$\sum_{i \geq n} \mu_i \geq \sum_{i \geq n} \nu_i, \quad \forall n \in \mathbb{Z}_{\geq 0}.$$

□

A very useful property of stochastic dominance is the following:

Lemma 7 For any probability distributions μ and ν on $\mathbb{Z}_{\geq 0}$ such that $\mu \succeq_s \nu$ and for any increasing function $f: \mathbb{Z}_{\geq 0} \rightarrow \mathbb{R}$,

$$\sum_{n=0}^{\infty} f(n) \mu_n \geq \sum_{n=0}^{\infty} f(n) \nu_n.$$

□

This is a standard result. See, for example, [30, Lemma 4.7.2].

To prove Proposition 1, we extend the notion of stochastic dominance to distributions defined over \mathbb{Z} .

Definition 8 (Reflected stochastic dominance) Let μ and ν be two probability distributions defined over \mathbb{Z} . Then μ is said to dominate ν in the sense of reflected stochastic dominance, which is denoted by $\mu \succeq_r \nu$, if

$$\sum_{i \geq n} (\mu_i + \mu_{-i}) \geq \sum_{i \geq n} (\nu_i + \nu_{-i}), \quad \forall n \in \mathbb{Z}_{> 0}.$$

□

Lemma 8 For any probability distributions μ and ν defined over \mathbb{Z} such that $\mu \succeq_r \nu$ and for any function $f: \mathbb{Z} \rightarrow \mathbb{R}$ that is even and increasing on $\mathbb{Z}_{\geq 0}$,

$$\sum_{n=-\infty}^{\infty} f(n) \mu_n \geq \sum_{n=-\infty}^{\infty} f(n) \nu_n.$$

□

Proof: Define distributions $\tilde{\mu}$ and $\tilde{\nu}$ over $\mathbb{Z}_{\geq 0}$ as follows: for every $n \in \mathbb{Z}_{\geq 0}$

$$\tilde{\mu}_n = \begin{cases} \mu_0, & \text{if } n = 0 \\ \mu_n + \mu_{-n}, & \text{otherwise;} \end{cases}$$

and $\tilde{\nu}$ defined similarly. An immediate consequence of the definitions is that

$$\mu \succeq_r \nu \implies \tilde{\mu} \succeq_s \tilde{\nu}. \quad (35)$$

For any even function $f: \mathbb{Z} \rightarrow \mathbb{R}$

$$\sum_{n=-\infty}^{\infty} f(n) \mu_n = \sum_{n=0}^{\infty} f(n) \tilde{\mu}_n. \quad (36)$$

The result follows from (35), (36), and Lemma 7. ■

Lemma 9 For any $e \in \mathbb{Z}_{\geq 0}$, $[P]_{e+1} \succeq_r [P]_e$, where $[P]_e$ denotes row e of P . □

Proof: To prove the result, we have to show that for any $n \in \mathbb{Z}_{\geq 0}$

$$\sum_{i \geq n+1} (P_{(e+1)i} + P_{(e+1)(-i)}) \geq \sum_{i \geq n+1} (P_{ei} + P_{e(-i)}),$$

or, equivalently,

$$\sum_{i=-n}^n P_{ei} \geq \sum_{i=-n}^n P_{(e+1)i}.$$

To prove the above, it is sufficient to show that

$$P_{ei} \geq P_{(e+1)(-i)}, \quad \forall e, i \in \mathbb{Z}_{\geq 0}. \quad (37)$$

Recall that $P_{ij} = p_{|i-j|}$ where $\{p_n\}_{n=0}^\infty$ is a decreasing sequence. Thus, $P_{ei} = p_{|e-i|}$ and $P_{(e+1)(-i)} = p_{e+i+1}$. Since e and i are positive, by the triangle inequality we have that $|e-i| \leq e+i < e+i+1$. Hence, $p_{|e-i|} \leq p_{e+i+1}$, which proves (37). ■

Finally, note the following obvious properties of even and increasing functions that we state without proof. Let EI denote ‘even and increasing on $\mathbb{Z}_{\geq 0}$ ’. Then

(P1) Sum of two EI functions is EI.

(P2) Pointwise minimum of two EI functions is EI.

We now prove Proposition 1.

Proof of Proposition 1: We prove the result by backward induction. The result is trivially true for V_T , which is the basis of induction. Assume that $V_{t+1}(\cdot; \lambda)$ is even and increasing on $\mathbb{Z}_{\geq 0}$. Define

$$\hat{V}_t(e; \lambda) = \sum_{n=-\infty}^{\infty} P_{en} V_{t+1}(n; \lambda).$$

We show that $\hat{V}_t(\cdot; \lambda)$ is even and increasing on $\mathbb{Z}_{\geq 0}$.

1) Consider

$$\begin{aligned} \hat{V}_t(-e; \lambda) &= \sum_{n=-\infty}^{\infty} P_{(-e)n} V_{t+1}(n; \lambda) \\ &= \sum_{-n=-\infty}^{\infty} P_{(-e)(-n)} V_{t+1}(-n; \lambda) \\ &\stackrel{(a)}{=} \sum_{n=-\infty}^{\infty} P_{en} V_{t+1}(n; \lambda) \\ &= \hat{V}_t(e; \lambda) \end{aligned}$$

where (a) uses $P_{en} = P_{(-e)(-n)}$ and $V_{t+1}(n; \lambda) = V_{t+1}(-n; \lambda)$. Hence, $\hat{V}_t(\cdot; \lambda)$ is even.

2) By Lemma 9, for all $e \in \mathbb{Z}_{\geq 0}$, $[P]_{e+1} \succeq_r [P]_e$. Since $V_{t+1}(\cdot; \lambda)$ is even and increasing on $\mathbb{Z}_{\geq 0}$, by Lemma 8

$$\hat{V}_t(e+1; \lambda) \geq \hat{V}_t(e; \lambda).$$

Hence, $\hat{V}_t(\cdot; \lambda)$ is increasing on $\mathbb{Z}_{\geq 0}$.

Now, V_t is given by

$$V_t(e; \lambda) = \min \{ \lambda + \hat{V}_t(0; \lambda), d(e) + \hat{V}_t(e; \lambda) \}.$$

By Assumption (A2), $d(\cdot)$ is even and increasing on $\mathbb{Z}_{\geq 0}$. Therefore, by properties (P1) and (P2) given above, the function $V_t(\cdot; \lambda)$ is even and increasing on $\mathbb{Z}_{\geq 0}$. This completes the induction step. Therefore, the result of Proposition 1 follows from the principle of induction. ■

APPENDIX C PROOF OF PROPOSITION 3

To prove the result, we introduce the notion of z -standard strategy from [31].

Definition 9 Consider a Markov chain with state space \mathcal{X} and a cost function $c : \mathcal{X} \rightarrow \mathbb{R}$. For $i, j \in \mathcal{X}$, let m_{ij} and C_{ij} denote the expected time and expected cost of the first passage from i to j . The Markov chain is called z -standard, $z \in \mathcal{X}$, if $m_{iz} < \infty$ and $C_{iz} < \infty$ for all $i \in \mathcal{X}$. □

Definition 10 (z -standard strategy) Let g be a (possibly randomized) stationary strategy for a Markov decision process. Then g is a z -standard strategy if the Markov chain induced by g is z -standard. □

We use the following result from [31, Proposition 7.5.3].

Proposition 9 If there exists a z -standard strategy for a Markov decision process, then the SEN conditions (S1) and (S2) hold for the reference state z . □

Lemma 10 *In the model considered in this paper, the strategy $f^{(0)}$ is 0-standard.* \square

Proof: The strategy $f^{(0)}$ is a ‘always transmit’ strategy. For any starting state e , the first passage time to 0 is $m_{e0} = 1 < \infty$ and the corresponding cost is $C_{e0} = \lambda < \infty$. Hence, $f^{(0)}$ is 0-standard. \blacksquare

Proof of Proposition 3: By Lemma 10 and Proposition 9, (S1) and (S2) hold in our model. From Proposition 2, we have that $V_\beta(e; \lambda) \geq V_\beta(0; \lambda)$. Hence, (S3) holds for $L_\lambda = 0$. \blacksquare

APPENDIX D PROOF OF PROPOSITION 4

We first consider the case $k = 0$. In this case, the recursive definition of $D_\beta^{(k)}$ and $N_\beta^{(k)}$, given by (9) and (10), simplify to the following:

$$D_\beta^{(0)}(e) = \beta \sum_{n=-\infty}^{\infty} P_{0n} D_\beta^{(0)}(n);$$

and

$$N_\beta^{(0)}(e) = (1 - \beta) + \beta \sum_{n=-\infty}^{\infty} P_{0n} N_\beta^{(0)}(n).$$

It can be easily verified that $D_\beta^{(0)}(e) = 0$ and $N_\beta^{(0)}(e) = 1$, $e \in \mathbb{Z}$, satisfy the above equations. From (11), we get that $C_\beta^{(0)}(e; \lambda) = \lambda$. This proves the first part of the proposition.

For $k > 0$, define

$$\begin{aligned} \hat{D}_\beta^{(k)} &:= \sum_{n=-\infty}^{\infty} P_{0n} D_\beta^{(k)}(n) \\ \hat{N}_\beta^{(k)} &:= \sum_{n=-\infty}^{\infty} P_{0n} N_\beta^{(k)}(n) \end{aligned}$$

From (9) and (10), we have that

$$D_\beta^{(k)}(0) = \beta \hat{D}_\beta^{(k)} \quad \text{and} \quad N_\beta^{(k)}(0) = \beta \hat{N}_\beta^{(k)}. \quad (38)$$

An equivalent representation of $D_\beta^{(k)}(0)$ is

$$D_\beta^{(k)}(0) = \mathbb{E} \left[(1 - \beta) \sum_{t=0}^{\tau^{(k)}-1} \beta^t d(E_t) + \beta^{\tau^{(k)}} [\beta \hat{D}_\beta^{(k)}] \mid E_0 = 0 \right]. \quad (39)$$

Using the strong Markov property and by substituting (13) and (38) in (39), we get that

$$D_\beta^{(k)}(0) = (1 - \beta) L_\beta^{(k)} + [1 - (1 - \beta) M_\beta^{(k)}] D_\beta^{(k)}(0).$$

Rearranging, we get that

$$D_\beta^{(k)}(0) = \frac{L_\beta^{(k)}}{M_\beta^{(k)}}.$$

Similarly, an equivalent representation of $N_\beta^{(k)}$ is

$$N_\beta^{(k)}(0) = \mathbb{E} \left[\beta^{\tau^{(k)}} [(1 - \beta) + \beta \hat{N}_\beta^{(k)}] \mid E_0 = 0 \right]. \quad (40)$$

Using the strong Markov property and by substituting (14) and (38) in (40), we get that

$$N_\beta^{(k)}(0) = [1 - (1 - \beta) M_\beta^{(k)}] [(1 - \beta) + N_\beta^{(k)}(0)].$$

Rearranging, we get that

$$N_\beta^{(k)}(0) = \frac{1}{M_\beta^{(k)}} - (1 - \beta).$$

The expression for $C_\beta^{(k)}(0; \lambda)$ follows from (11).

APPENDIX E
PROOF OF PROPOSITION 5

A. Analytic expressions of $L_\beta^{(k)}$ and $M_\beta^{(k)}$

For a matrix A , let $[A]_0$ denote the row with index 0. (Recall that our index set includes negative values as well). $P^{(k)}$ is a sub-stochastic matrix that captures the probability of the Markov chain not leaving the set $S^{(k)}$. Therefore,

$$\begin{aligned}
L_\beta^{(k)} &:= \mathbb{E} \left[\sum_{t=0}^{\tau^{(k)}-1} \beta^t d(E_t) \mid E_0 = 0 \right] \\
&= \sum_{t=0}^{\infty} \beta^t \left[\sum_{e \in S^{(k)}} (P_{0e}^{(k)})^t d(e) \right] \\
&= \sum_{t=0}^{\infty} \left\langle [(\beta P^{(k)})^t]_0, d^{(k)} \right\rangle \\
&= \left\langle \sum_{t=0}^{\infty} [(\beta P^{(k)})^t]_0, d^{(k)} \right\rangle \\
&= \left\langle \left[\sum_{t=0}^{\infty} (\beta P^{(k)})^t \right]_0, d^{(k)} \right\rangle \\
&= \langle [Q_\beta^{(k)}]_0, d^{(k)} \rangle
\end{aligned} \tag{41}$$

where we used the fact that, since $P^{(k)}$ is a sub-stochastic matrix, we have

$$Q_\beta^{(k)} = \sum_{t=0}^{\infty} \beta^t (P^{(k)})^t. \tag{42}$$

To prove (19), note that $M_\beta^{(k)}$ may also be written as

$$M_\beta^{(k)} = \mathbb{E} \left[\sum_{t=0}^{\tau^{(k)}-1} \beta^t \mid E_0 = 0 \right].$$

The rest of the proof is along the same lines as (41).

B. Monotonicity of $L_\beta^{(k)}$ and $M_\beta^{(k)}$

To prove the monotonicity of $L_\beta^{(k)}$ and $M_\beta^{(k)}$, we use the following recursive expression for $P^{(k)}$.

Lemma 11 For any $t \in \mathbb{Z}_{\geq 0}$, $(P^{(k+1)})^t$ is of the form

$$(P^{(k+1)})^t = \begin{bmatrix} a_t^{(k)} & b_t^{(k)} & c_t^{(k)} \\ (b_t^{(k)})^\top & (P^{(k)})^t + A_t^{(k)} & (d_t^{(k)})^\top \\ c_t^{(k)} & d_t^{(k)} & a_t^{(k)} \end{bmatrix}, \tag{43}$$

where $a_t^{(k)}$ and $c_t^{(k)}$ are positive scalars; $A_t^{(k)}$ is a $2k-1 \times 2k-1$ dimensional matrix with all positive elements for all $t \in \mathbb{Z}_{>1}$ and with $A_1^{(k)} = 0_{2k-1 \times 2k-1}$; $b_t^{(k)}$, $d_t^{(k)}$ are the vectors of dimension $1 \times 2k-1$ with all positive elements; they are all given by the following recursive equations

$$a_0^{(k)} = 1, \quad c_0^{(k)} = 0, \quad b_0^{(k)} = 0_{1 \times 2k-1} = d_0^{(k)},$$

and for $t > 1$

$$\begin{aligned}
a_t^{(k)} &= a_{t-1}^{(k)} a_1^{(k)} + b_{t-1}^{(k)} (b_1^{(k)})^\top + c_{t-1}^{(k)} c_1^{(k)} \\
&= c_{t-1}^{(k)} c_1^{(k)} + d_{t-1}^{(k)} (d_1^{(k)})^\top + a_{t-1}^{(k)} a_1^{(k)},
\end{aligned}$$

$$\begin{aligned}
c_t^{(k)} &= a_{t-1}^{(k)} c_1^{(k)} + b_{t-1}^{(k)} (d_1^{(k)})^\top + c_{t-1}^{(k)} a_1^{(k)} \\
&= c_{t-1}^{(k)} a_1^{(k)} + d_{t-1}^{(k)} (b_1^{(k)})^\top + a_{t-1}^{(k)} c_1^{(k)}, \\
b_t^{(k)} &= a_{t-1}^{(k)} b_1^{(k)} + b_{t-1}^{(k)} P^{(k)} + c_{t-1}^{(k)} d_1^{(k)}, \\
d_t^{(k)} &= c_{t-1}^{(k)} b_1^{(k)} + d_{t-1}^{(k)} P^{(k)} + a_{t-1}^{(k)} d_1^{(k)}, \\
A_t^{(k)} &= (b_{t-1}^{(k)})^\top b_1^{(k)} + A_{t-1}^{(k)} P^{(k)} + (d_{t-1}^{(k)})^\top d_1^{(k)}.
\end{aligned}$$

□

Using the above recursion and the fact that all elements of vectors $b_t^{(k)}$, $d_t^{(k)}$, and matrix $A_t^{(k)}$ are positive, we get the following.

Lemma 12 For all $t \in \mathbb{Z}_{\geq 0}$,

$$\langle [(P^{(k)})^t]_0, d^{(k)} \rangle < \langle [(P^{(k+1)})^t]_0, d^{(k+1)} \rangle,$$

and

$$\langle [(P^{(k)})^t]_0, \mathbf{1}_{2k-1} \rangle < \langle [(P^{(k+1)})^t]_0, \mathbf{1}_{2k+1} \rangle.$$

□

Now, to prove the monotonicity of $L_\beta^{(k)}$, consider

$$\begin{aligned}
L_\beta^{(k)} &= \langle [Q_\beta^{(k)}]_0, d^{(k)} \rangle \\
&= \left\langle \sum_{t=0}^{\infty} [\beta^t (P^{(k)})^t]_0, d^{(k)} \right\rangle \\
&\stackrel{(a)}{<} \left\langle \sum_{t=0}^{\infty} [\beta^t (P^{(k+1)})^t]_0, d^{(k+1)} \right\rangle \\
&= \langle [Q_\beta^{(k+1)}]_0, d^{(k+1)} \rangle = L_\beta^{(k+1)}
\end{aligned} \tag{44}$$

where (a) follows from Lemma 12. Monotonicity of $M_\beta^{(k)}$ can be proved along similar lines. This completes the proof of the proposition.

Proof of Lemma 11: We prove the result by induction. For $t = 1$, we have that

$$\begin{aligned}
P^{(k+1)} &= \begin{bmatrix} a_1^{(k)} & b_1^{(k)} & c_1^{(k)} \\ (b_1^{(k)})^\top & P^{(k)} & (d_1^{(k)})^\top \\ c_1^{(k)} & d_1^{(k)} & a_1^{(k)} \end{bmatrix} \\
&= \begin{bmatrix} a_1^{(k)} & b_1^{(k)} & c_1^{(k)} \\ (b_1^{(k)})^\top & P^{(k)} + A_1^{(k)} & (d_1^{(k)})^\top \\ c_1^{(k)} & d_1^{(k)} & a_1^{(k)} \end{bmatrix}
\end{aligned}$$

where $a_1^{(k)} = P_{00}$, $b_1^{(k)} = [P_{01} \cdots P_{0k}]$, $c_1^{(k)} = P_{02(k-1)}$ and $d_1^{(k)} = [P_{0k} \cdots P_{01}]$. Hence the result holds for $t = 1$.

Now, assume that the result is true for $t-1$ for some $t > 1$. Note that $(P^{(k+1)})^{t-1}$ is symmetric because any positive power of a symmetric matrix is symmetric. Hence, $(P^{(k+1)})^t$ is a symmetric matrix given by

$$\begin{aligned}
&(P^{(k+1)})^t \\
&= (P^{(k+1)})^{t-1} P^{(k+1)} \\
&= \begin{bmatrix} a_{t-1}^{(k)} & b_{t-1}^{(k)} & c_{t-1}^{(k)} \\ (b_{t-1}^{(k)})^\top & (P^{(k)})^{t-1} + A_{t-1}^{(k)} & (d_{t-1}^{(k)})^\top \\ c_{t-1}^{(k)} & d_{t-1}^{(k)} & a_{t-1}^{(k)} \end{bmatrix} \begin{bmatrix} a_1^{(k)} & b_1^{(k)} & c_1^{(k)} \\ (b_1^{(k)})^\top & P^{(k)} & (d_1^{(k)})^\top \\ c_1^{(k)} & d_1^{(k)} & a_1^{(k)} \end{bmatrix} \\
&= \begin{bmatrix} a_t^{(k)} & b_t^{(k)} & c_t^{(k)} \\ (b_t^{(k)})^\top & (P^{(k)})^t + A_t^{(k)} & (d_t^{(k)})^\top \\ c_t^{(k)} & d_t^{(k)} & a_t^{(k)} \end{bmatrix}
\end{aligned} \tag{45}$$

The recursive expressions in Lemma 11 follow from comparing corresponding terms in (45). Hence, by the principle of induction, the result is true for all t . ■

Proof of Lemma 12: We only prove the first inequality. The second inequality can be proved along similar lines.

By Lemma 11, we have

$$\begin{aligned}
& \langle [(P^{(k+1)})^t]_0, d^{(k+1)} \rangle \\
&= \left[(b_t^{(k)})^\top \quad (P^{(k)})^t + A_t^{(k)} \quad (d_t^{(k)})^\top \right]_0 \begin{bmatrix} d(-k-1) \\ d^{(k)} \\ d(k+1) \end{bmatrix} \\
&< \langle [(P^{(k)})^t]_0, d^{(k)} \rangle
\end{aligned}$$

where the last inequality follows from the fact that all elements of the vectors $\mathbf{b}_t^{(k)}$, $\mathbf{d}_t^{(k)}$ and the matrix $A_t^{(k)}$ are positive. ■

C. Monotonicity of $D_\beta^{(k)}$

Lastly, we prove the monotonicity of $D_\beta^{(k)}(e)$ in k .

Define the operator $T^{(k+1)} : (\mathbb{Z} \rightarrow \mathbb{R}) \rightarrow (\mathbb{Z} \rightarrow \mathbb{R})$ as follows. For any $D : \mathbb{Z} \rightarrow \mathbb{R}$,

$$[T^{(k+1)}D](e) = \begin{cases} \beta \sum_{n=-\infty}^{\infty} P_{0n} D(n; \lambda), & |e| \geq k+1 \\ (1-\beta)d(e) + \beta \sum_{n=-\infty}^{\infty} P_{en} D(n; \lambda), & |e| < k+1. \end{cases} \quad (46)$$

Note that, as a consequence of Theorem 3, the operator $T^{(k+1)}$, $k \in \mathbb{Z}_{>0}$, is a contraction and $D_\beta^{(k+1)}$ is its a unique bounded fixed point. Next, define function $D_\beta^{(k,m)}$, $m \in \mathbb{Z}_{\geq 0}$, as follows:

$$\begin{aligned}
D_\beta^{(k,0)} &= D_\beta^{(k)}, \\
D_\beta^{(k,m)} &= T^{(k+1)} D_\beta^{(k,m-1)}, \quad m \in \mathbb{Z}_{>0}.
\end{aligned} \quad (47)$$

Let $p_k = P_{0k}$, $k \in \mathbb{Z}_{\geq 0}$ and $b := \sup\{k \in \mathbb{Z}_{\geq 0} \mid p_k > 0\}$ and define

$$\begin{aligned}
A_+^{(m)} &= \{k, k-1, \dots, \max(k-mb, 0)\}, \\
A_-^{(m)} &= \{-k, -k+1, \dots, \min(-k+mb, 0)\}
\end{aligned}$$

and

$$A^{(m)} = A_+^{(m)} \cup A_-^{(m)}.$$

Note that $A^{(0)} = \{-k, k\}$ and

$$A^{(m)} \subseteq A^{(m+1)} \subseteq \{-k, \dots, k\}.$$

Let m° be the smallest integer such that $A^{(m^\circ)} = \{-k, \dots, k\}$ (in particular, if $b = \infty$, then $m^\circ = 2$). We will show the following

Lemma 13 For any $m \in \{0, 1, \dots, m^\circ\}$

$$D_\beta^{(k,m+1)}(e) > D_\beta^{(k)}(e), \quad \forall e \in A^{(m)}$$

and

$$D_\beta^{(k,m+1)}(e) \geq D_\beta^{(k)}(e), \quad \forall e \notin A^{(m)}.$$

□

Next, define

$$\begin{aligned}
B_+^{(m)} &= \{k+1, \dots, k+mb\}, \\
B_-^{(m)} &= \{-k-1, \dots, -k-mb\}
\end{aligned}$$

and

$$B^{(m)} = B_+^{(m)} \cup B_-^{(m)}, \quad B^{(0)} = \phi.$$

We will also show that

Lemma 14 For $m \in \mathbb{Z}_{\geq 0}$,

$$D_\beta^{(k,m+m^\circ+1)}(e) > D_\beta^{(k)}(e), \quad \forall e \in B^{(m)} \cup A^{(m^\circ)}$$

and

$$D_{\beta}^{(k,m+m^{\circ}+1)}(e) \geq D_{\beta}^{(k)}(e), \quad \forall e \notin B^{(m)} \cup A^{(m^{\circ})}. \quad \square$$

Recall that $T^{(k+1)}$ is a contraction operator with $D^{(k+1)}$ as its fixed point. Since $\lim_{m \rightarrow \infty} B^{(m)} \cup A^{(m^{\circ})} = \mathbb{Z}$, we have that

$$\begin{aligned} D_{\beta}^{(k+1)}(e) &= \lim_{m \rightarrow \infty} D_{\beta}^{(k,m+m^{\circ}+1)}(e) \\ &> D_{\beta}^{(k)}(e), \quad \forall e \in \mathbb{Z}. \end{aligned}$$

Proof of Lemma 13: We prove the result by induction. Consider $m = 0$. Analogous to Proposition 2, we can show that $D_{\beta}^{(k)}(e)$ is even and increasing in e . By Lemma 8 and 9, $P_{en} \succeq_r P_{0n}$. Hence,

$$\sum_{n=-\infty}^{\infty} P_{en} D_{\beta}^{(k)}(n) \geq \sum_{n=-\infty}^{\infty} P_{0n} D_{\beta}^{(k)}(n). \quad (48)$$

For $e \in A^{(0)} = \{-k, k\}$,

$$D_{\beta}^{(k,1)}(e) = (1 - \beta)d(e) + \beta \sum_{n=-\infty}^{\infty} P_{en} D_{\beta}^{(k)}(n) \quad (49)$$

and

$$D_{\beta}^{(k)}(e) = \beta \sum_{n=-\infty}^{\infty} P_{0n} D_{\beta}^{(k)}(n). \quad (50)$$

By (48) and by Assumption (A2b),

$$D_{\beta}^{(k,1)}(e) > D_{\beta}^{(k)}(e), \quad \forall e \in A^{(0)} \quad (51)$$

$$D_{\beta}^{(k,1)}(e) \stackrel{(a)}{=} D_{\beta}^{(k)}(e), \quad \forall e \notin A^{(0)}, \quad (52)$$

where the equality (a) holds since both sides have same expressions. Now, we show the result for $m = 1$. Pick any arbitrary $e \in A^{(1)}$. We have from (46)

$$D_{\beta}^{(k,2)}(e) = (1 - \beta)d(e) + \beta \sum_{n=-\infty}^{\infty} P_{en} D_{\beta}^{(k,1)}(n), \quad (53)$$

Furthermore, from (9), we have

$$D_{\beta}^{(k)}(e) = \begin{cases} (1 - \beta)d(e) + \beta \sum_{n=-\infty}^{\infty} P_{en} D_{\beta}^{(k)}(n), & e \in A^{(1)} \setminus A^{(0)} \\ \beta \sum_{n=-\infty}^{\infty} P_{0n} D_{\beta}^{(k)}(n), & e \in A^{(0)}. \end{cases} \quad (54)$$

Since $e \in A^{(1)}$, $P_{ek} > 0$ and $P_{e(-k)} > 0$. Hence, by (49)–(50),

$$P_{ek} D_{\beta}^{(k,1)}(n) > P_{ek} D_{\beta}^{(k)}(n), \quad n \in \{-k, k\}.$$

Combining the above with (51)–(52), we get

$$\sum_{n=-\infty}^{\infty} P_{en} D_{\beta}^{(k,1)}(n) > \sum_{n=-\infty}^{\infty} P_{en} D_{\beta}^{(k)}(n), \quad \forall e \in A^{(1)},$$

and hence, by (53) and (54),

$$D_{\beta}^{(k,2)}(e) > D_{\beta}^{(k)}(e), \quad \forall e \in A^{(1)} \setminus A^{(0)}. \quad (55)$$

Also, by (51) and using monotonicity of $T^{(k+1)}$, we get

$$D_{\beta}^{(k,2)}(e) \geq D_{\beta}^{(k,1)}(e) > D_{\beta}^{(k)}(e), \quad \forall e \in A^{(0)}. \quad (56)$$

Combining (55) and (56), we get that

$$D_{\beta}^{(k,2)}(e) > D_{\beta}^{(k)}(e), \quad \forall e \in A^{(1)}.$$

Furthermore, since $D_{\beta}^{(k,1)}(e) \geq D_{\beta}^{(k)}(e)$, $\forall e \in \mathbb{Z}$, by monotonicity of $T^{(k+1)}$,

$$D_{\beta}^{(k,2)}(e) \geq D_{\beta}^{(k,1)}(e) \geq D_{\beta}^{(k)}(e), \quad \forall e \in \mathbb{Z}.$$

Now, suppose the result of Lemma 13 is true for some $(m-1)$, where $0 < m < m^\circ$. For any $e \in A^{(m)}$

$$D_\beta^{(k,m+1)}(e) = (1-\beta)d(e) + \beta \sum_{n=-\infty}^{\infty} P_{en} D_\beta^{(k,m)}(n), \quad (57)$$

$$D_\beta^{(k)}(e) = \begin{cases} (1-\beta)d(e) + \beta \sum_{n=-\infty}^{\infty} P_{en} D_\beta^{(k)}(n), & e \in A^{(m)} \setminus A^{(0)} \\ \beta \sum_{n=-\infty}^{\infty} P_{0n} D_\beta^{(k)}(n), & e \in A^{(0)}. \end{cases} \quad (58)$$

Consider any $e \in A_+^{(m)}$. If $e \in A_+^{(m-1)}$, then by monotonicity of $T^{(k+1)}$,

$$D_\beta^{(k,m+1)}(e) \geq D_\beta^{(k,m)}(e) > D_\beta^{(k)}(e), \quad (59)$$

where the last inequality follows from the induction hypothesis. If $e \notin A_+^{(m-1)}$, then

- $(e+b) \in A_+^{(m-1)} \implies D_\beta^{(k,m)}(e+b) > D_\beta^{(k)}(e+b),$
- $P_{e(e+b)} = P_{0b} > 0 \implies P_{e(e+b)} D_\beta^{(k,m)}(e+b) > P_{e(e+b)} D_\beta^{(k)}(e+b).$

Therefore,

$$\sum_{n=-\infty}^{\infty} P_{en} D_\beta^{(k,m)}(n) > \sum_{n=-\infty}^{\infty} P_{en} D_\beta^{(k)}(n). \quad (60)$$

Combining (57), (58), (59) and (60), we get

$$D_\beta^{(k,m+1)}(e) > D_\beta^{(k)}(e), \quad \forall e \in A_+^{(m)} \setminus A^{(0)}. \quad (61)$$

Furthermore, by (52) and monotonicity of $T^{(k+1)}$, we have

$$D_\beta^{(k,m+1)}(e) \geq D_\beta^{(k,1)}(e) > D_\beta^{(k)}(e), \quad \forall e \in A^{(0)}. \quad (62)$$

Combining (61) and (62), we get that

$$D_\beta^{(k,m+1)}(e) > D_\beta^{(k)}(e), \quad \forall e \in A_+^{(m)}.$$

Using a similar argument as above, we can also show that the above inequality holds for $e \in A_-^{(m)}$. Also, by monotonicity of $T^{(k+1)}$, we have that $D_\beta^{(k,m+m^\circ+1)}(e) \geq D_\beta^{(k,m+m^\circ)}(e) \geq D_\beta^{(k)}(e)$, $\forall e \in \mathbb{Z}$. This completes the induction step.

Hence, by principle of induction, Lemma 13 is true. ■

Proof of Lemma 14: We prove the result using induction. It is easy to see that by Lemma 13, the statements of the lemma are true for $m = 0$. For $m = 1$, note that $B^{(m-1)} = \phi$ and hence $B^{(m-1)} \cup A^{(m^\circ)} = A^{(m^\circ)}$. By monotonicity of $T^{(k+1)}$ and Lemma 13, we have the following:

$$D_\beta^{(k,m^\circ+2)} \geq D_\beta^{(k,m^\circ+1)} > D_\beta^{(k)}, \quad \forall e \in A^{(m^\circ)},$$

which is the result of the lemma for $m = 1$. Now, let us assume that Lemma 14 is true for some integer $m > 1$, i.e.

$$D_\beta^{(k,m+m^\circ)}(e) > D_\beta^{(k)}(e), \quad \forall e \in B^{(m-1)} \cup A^{(m^\circ)}. \quad (63)$$

Now, consider $e \in B_+^{(m)}$. If $e \in B_+^{(m-1)}$, then by monotonicity of $T^{(k+1)}$,

$$D_\beta^{(k,m+m^\circ+1)}(e) \geq D_\beta^{(k,m+m^\circ)}(e) > D_\beta^{(k)}(e),$$

where the last inequality follows from the induction hypothesis. If $e \notin B_+^{(m-1)}$, then

- $(e-b) \in B_+^{(m-1)} \implies D_\beta^{(k,m+m^\circ)}(e-b) > D_\beta^{(k)}(e-b),$
- $P_{e(e-b)} = P_{0(-b)} = P_{0b} > 0 \implies P_{e(e-b)} D_\beta^{(k,m+m^\circ)}(e-b) > P_{e(e-b)} D_\beta^{(k)}(e-b).$

Thus,

$$\sum_{n=-\infty}^{\infty} P_{en} D_\beta^{(k,m+m^\circ)}(n) > \sum_{n=-\infty}^{\infty} P_{en} D_\beta^{(k)}(n). \quad (64)$$

Combining (48), (49) and (64) we get

$$D_\beta^{(k,m+m^\circ+1)}(e) > D_\beta^{(k)}(e), \quad \forall e \in B_+^{(m)} \cup A^{(m^\circ)}.$$

Proceeding in a similar way as above, it can be shown that the above inequality holds for all $e \in B_-^{(m)} \cup A^{(m^\circ)}$. Also, by monotonicity of $T^{(k+1)}$, we have that $D_\beta^{(k,m+m^\circ+1)}(e) \geq D_\beta^{(k,m+m^\circ)}(e) \geq D_\beta^{(k)}(e)$, $\forall e \in \mathbb{Z}$. This completes the induction step. Hence, by principle of induction, Lemma 14 is true. ■

APPENDIX F
PROOF OF THEOREM 5

A. Proof of part 1)

Consider

$$\begin{aligned}
& C_\beta^{(k)}(0; \lambda) - C_\beta^{(k+1)}(0; \lambda) \\
& \stackrel{(a)}{=} [C_\beta^{(k)}(0; \lambda) - C_\beta^{(k)}(0; \lambda_\beta^{(k)})] \\
& \quad - [C_\beta^{(k+1)}(0; \lambda) - C_\beta^{(k+1)}(0; \lambda_\beta^{(k)})] \\
& \stackrel{(b)}{=} [\lambda - \lambda_\beta^{(k)}] \left[\frac{1}{M_\beta^{(k)}} - \frac{1}{M_\beta^{(k+1)}} \right]
\end{aligned} \tag{65}$$

where (a) follows from (22) and (b) follows from Proposition 4. By Proposition 5, $M_\beta^{(k)} < M_\beta^{(k+1)}$; hence, the sign of $C_\beta^{(k)}(0; \lambda) - C_\beta^{(k+1)}(0; \lambda)$ is the same as that of $(\lambda - \lambda_\beta^{(k)})$.

Now, consider a $\lambda \in (\lambda_\beta^{(k)}, \lambda_\beta^{(k+1)})$. By Assumption (A4), for any $m \in \mathbb{Z}_{\geq 0}$ such that $m \leq k$, $\lambda_\beta^{(m)} \leq \lambda$. Hence, by (65)

$$C_\beta^{(m)}(0; \lambda) \geq C_\beta^{(m+1)}(0; \lambda), \quad \forall m \leq k. \tag{66}$$

Similarly, for any $m \in \mathbb{Z}_{\geq 0}$ such that $m \geq k+1$, $\lambda_\beta^{(m)} \geq \lambda$. Hence, by (65)

$$C_\beta^{(m+1)}(0; \lambda) \geq C_\beta^{(m)}(0; \lambda), \quad \forall m \geq k+1. \tag{67}$$

Combining (66) and (67), we get that $f^{(k+1)}$ is optimal among all threshold strategies; and, by Theorem 3, is also globally optimum.

B. Proof of part 2)

By the previous part and Proposition 4, for any $\lambda \in (\lambda_\beta^{(k)}, \lambda_\beta^{(k+1)})$,

$$C_\beta^*(\lambda) = V_\beta(0; \lambda) = \frac{L_\beta^{(k)} + \lambda}{M_\beta^{(k)}} - \lambda(1 - \beta),$$

which is continuous and linear in λ . Thus, $V_\beta(0; \lambda)$ is piecewise linear in λ .

Moreover,

$$\lim_{\lambda \downarrow \lambda_\beta^{(k)}} V_\beta(0; \lambda) = C_\beta^{(k+1)}(0; \lambda_\beta^{(k)});$$

and

$$\lim_{\lambda \uparrow \lambda_\beta^{(k)}} V_\beta(0; \lambda) = C_\beta^{(k)}(0; \lambda_\beta^{(k)}).$$

By (22), both these terms are equal. Therefore, $V_\beta(0; \lambda)$ is continuous.

Next, note that the slope of $V_\beta(0; \lambda)$ in the interval $(\lambda_\beta^{(k)}, \lambda_\beta^{(k+1)})$, is given by

$$\frac{C_\beta^{(k+1)}(0; \lambda_\beta^{(k+1)}) - C_\beta^{(k+1)}(0; \lambda_\beta^{(k)})}{\lambda_\beta^{(k+1)} - \lambda_\beta^{(k)}} = \frac{1}{M_\beta^{(k+1)}} - (1 - \beta),$$

where we have used the result of Proposition 4. By Proposition 5, $\{M_\beta^{(k)}\}_{k=0}^\infty$ is a strictly increasing sequence. Therefore, the slope of $V_\beta(0; \lambda)$ decreases as λ increases. Hence, $V_\beta(0; \lambda)$ is concave.

Finally, from (14), we get that $M_\beta^{(k+1)} \leq 1/(1 - \beta)$. Hence, the slope of $V_\beta(0; \lambda)$ calculated above is always non-negative. Hence, $V_\beta(0; \lambda)$ is increasing in λ .

APPENDIX G
PROOF OF THEOREM 6

Consider a $\lambda \in (\lambda_1^{(k)}, \lambda_1^{(k+1)})$. By definition, $\lambda_1^{(k)} = \lim_{\beta \uparrow 1} \lambda_\beta^{(k)}$. Therefore, there exists a $\beta^* \in (0, 1)$ such that for all $\beta \in (\beta^*, 1)$, $\lambda \in (\lambda_\beta^{(k)}, \lambda_\beta^{(k+1)})$. By Theorem 5, the strategy $f^{(k+1)}$ is discounted cost optimal for all $\beta \in (\beta^*, 1)$. Hence, by Theorem 4, the strategy $f^{(k+1)}$ is also optimal for the long-term average cost setup and

$$C_1^*(\lambda) = \lim_{\beta \uparrow 1} V_\beta(0; \lambda) = \frac{L_1^{(k)} + \lambda}{M_1^{(k)}}.$$

By an argument similar to the proof of part 2) of Theorem 5, we can show that $C_1^*(\lambda)$ is piecewise linear, continuous, and concave (instead of Propositions 4 and 5, we would use Propositions 7 and 8. The rest of the argument remains the same).

REFERENCES

- [1] T. Linder and G. Lugosi, "A zero-delay sequential scheme for lossy coding of individual sequences," *IEEE Trans. Inf. Theory*, vol. 47, no. 6, pp. 2533–2538, 2001.
- [2] T. Weissman and N. Merhav, "On limited-delay lossy coding and filtering of individual sequences," *IEEE Trans. Inf. Theory*, vol. 48, no. 3, pp. 721–733, 2002.
- [3] A. György, T. Linder, and G. Lugosi, "Efficient adaptive algorithms and minimax bounds for zero-delay lossy source coding," *IEEE Trans. Signal Process.*, vol. 52, no. 8, pp. 2337–2347, 2004.
- [4] S. Matloub and T. Weissman, "Universal zero-delay joint source-channel coding," *IEEE Trans. Inf. Theory*, vol. 52, no. 12, pp. 5240–5250, Dec. 2006.
- [5] H. S. Witsenhausen, "On the structure of real-time source coders," *Bell System Technical Journal*, vol. 58, no. 6, pp. 1437–1451, July–August 1979.
- [6] J. C. Walrand and P. Varaiya, "Optimal causal coding-decoding problems," *IEEE Trans. Inf. Theory*, vol. 29, no. 6, pp. 814–820, Nov. 1983.
- [7] D. Teneketzis, "On the structure of optimal real-time encoders and decoders in noisy communication," *IEEE Trans. Inf. Theory*, pp. 4017–4035, Sep. 2006.
- [8] A. Mahajan and D. Teneketzis, "Optimal design of sequential real-time communication systems," *IEEE Trans. Inf. Theory*, vol. 55, no. 11, pp. 5317–5338, Nov. 2009.
- [9] Y. Kaspi and N. Merhav, "Structure theorems for real-time variable rate coding with and without side information," *IEEE Trans. Inf. Theory*, vol. 58, no. 12, pp. 7135–7153, 2012.
- [10] H. Asnani and T. Weissman, "Real-time coding with limited lookahead," *IEEE Trans. Inf. Theory*, vol. 59, no. 6, pp. 3582–3606, 2013.
- [11] O. C. Imer and T. Basar, "Optimal estimation with limited measurements," *Joint 44th IEEE Conference on Decision and Control and European Control Conference*, vol. 29, pp. 1029 – 1034, 2005.
- [12] Y. Xu and J. P. Hespanha, "Optimal communication logics in networked control systems," in *Proceedings of 43rd IEEE Conference on Decision and Control*, vol. 4, 2004, pp. 3527–3532.
- [13] G. M. Lipsa and N. Martins, "Remote state estimation with communication costs for first-order LTI systems," *IEEE Trans. Autom. Control*, vol. 56, no. 9, pp. 2013–2025, Sep. 2011.
- [14] A. Nayyar, T. Basar, D. Teneketzis, and V. Veeravalli, "Optimal strategies for communication and remote estimation with an energy harvesting sensor," *IEEE Trans. Autom. Control*, vol. 58, no. 9, pp. 2246–2260, 2013.
- [15] A. Molin and S. Hirche, "An iterative algorithm for optimal event-triggered estimation," in *4th IFAC Conference on Analysis and Design of Hybrid Systems (ADHS'12)*, 2012, pp. 64–69.
- [16] C. Rago, P. Willett, and Y. Bar-Shalom, "Censoring sensors: A low-communication rate scheme for distributed detection," *IEEE Transactions on Aerospace and Electronic Systems*, vol. 32, no. 2, pp. 554–568, April 1996.
- [17] S. Appadwedula, V. V. Veeravalli, and D. L. Jones, "Decentralized detection with censoring sensors," *IEEE Transactions on Signal Processing*, vol. 56, no. 4, pp. 1362–1373, April 2008.
- [18] M. Athans, "On the determination of optimal costly measurement strategies for linear stochastic systems," *Automatica*, vol. 8, no. 4, pp. 397–412, 1972.
- [19] J. Geromel, "Global optimization of measurement strategies for linear stochastic systems," *Automatica*, vol. 25, no. 2, pp. 293–300, 1989.
- [20] W. Wu, A. Araposthathis, and V. V. Veeravalli, "Optimal sensor querying: General Markovian and LQG models with controlled observations," *IEEE Transactions on Automatic Control*, vol. 53, no. 6, pp. 1392–1405, 2008.
- [21] D. Shuman and M. Liu, "Optimal sleep scheduling for a wireless sensor network node," in *Proceedings of the Asilomar Conference on Signals, Systems, and Computers*, October 2006, pp. 1337–1341.
- [22] M. Sarkar and R. L. Cruz, "Analysis of power management for energy and delay trade-off in a WLAN," in *Proceedings of the Conference on Information Sciences and Systems*, March 2004.
- [23] —, "An adaptive sleep algorithm for efficient power management in WLANs," in *Proceedings of the Vehicular Technology Conference*, May 2005.
- [24] A. Federgruen and K. C. So, "Optimality of threshold policies in single-server queueing systems with server vacations," *Adv. Appl. Prob.*, vol. 23, no. 2, pp. 388–405, 1991.
- [25] K. J. Åström, *Analysis and Design of Nonlinear Control Systems*. Berlin, Heidelberg: Springer, 2008, ch. Event based control.
- [26] M. Rabi, G. Moustakides, and J. Baras, "Adaptive sampling for linear state estimation," *SIAM Journal on Control and Optimization*, vol. 50, no. 2, pp. 672–702, 2012.
- [27] X. Meng and T. Chen, "Optimal sampling and performance comparison of periodic and event based impulse control," *IEEE Transactions of Automatic Control*, vol. 57, no. 12, pp. 3252–3259, 2012.
- [28] S. Yüksel and T. Başar, *Stochastic Networked Control Systems: Stabilization and Optimization under Information Constraints*. Boston, MA: Birkhäuser, 2013.
- [29] A. Mahajan, N. Martins, M. Rotkowitz, and S. Yüksel, "Information structures in optimal decentralized control," in *Proc. 51st IEEE Conf. Decision and Control*, Maui, Hawaii, Dec. 2012, pp. 1291 – 1306.
- [30] M. Puterman, *Markov decision processes: Discrete Stochastic Dynamic Programming*. John Wiley and Sons, 1994.
- [31] L. I. Sennott, *Stochastic dynamic programming and the control of queueing systems*. New York, NY, USA: Wiley, 1999.
- [32] P. J. Weeda, *Finite generalized Markov programming*, ser. Mathematical Centre tracts. Amsterdam: Mathematisch Centrum, 1979, vol. 92.
- [33] J. Gittins, K. Glazebrook, and R. Weber, *Multi-Armed Bandit Allocation Indices*, 2nd ed. John Wiley and Sons, Ltd, 2011.
- [34] L. I. Sennott, "Computing average optimal constrained policies in stochastic dynamic programming," *Probability in the Engineering and Information Sciences*, vol. 15, pp. 103–133, 2001.
- [35] V. Borkar, "A convex analytic approach to Markov decision processes," *Probability Theory and Related Fields*, vol. 78, no. 4, pp. 583–602, 1988. [Online]. Available: <http://dx.doi.org/10.1007/BF00353877>
- [36] D. Luenberger, *Optimization by Vector Space Methods*, ser. Professional Series. Wiley, 1968. [Online]. Available: <http://books.google.ca/books?id=IZU0CAH4RccC>
- [37] G. Hu and R. O'Connell, "Analytical inversion of symmetric tridiagonal matrices," *Journal of Physics A: Mathematical and General*, vol. 29, no. 7, p. 1511, 1996.
- [38] E. Feinberg, "Optimality of deterministic policies for certain stochastic control problems with multiple criteria and constraints," in *Mathematical Control Theory and Finance*, A. Sarychev, A. Shiryayev, M. Guerra, and M. Grossinho, Eds. Springer Berlin Heidelberg, 2008, pp. 137–148.
- [39] A. Schwartz and A. M. Makowski, "An optimal adaptive scheme for two competing queues with constraints," in *Analysis and optimization of systems*. Springer Berlin Heidelberg, 1986, pp. 515–532.
- [40] D.-J. Ma, A. M. Makowski, and A. Schwartz, "Stochastic approximations for finite-state Markov chains," *Stochastic Processes and Their Applications*, vol. 35, no. 1, pp. 27–45, 1990.
- [41] E. Altman and A. Schwartz, "Time-sharing policies for controlled Markov chains," *Operations Research*, vol. 41, no. 6, pp. 1116–1124, 1993.
- [42] B. Hajek, K. Mitzel, and S. Yang, "Paging and registration in cellular networks: Jointly optimal policies and an iterative algorithm," *IEEE Trans. Inf. Theory*, vol. 64, pp. 608–622, Feb. 2008.
- [43] L. Wang, J. Woo, and M. Madiman, "A lower bound on Rényi entropy of convolutions in the integers," in *Proceedings of the 2014 IEEE International Symposium on Information Theory*, Jul. 2014, pp. 2829–2833.